



Unit 7

AC losses in Superconductors

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Scope of the Lesson



- AC losses – general classification
 1. Hysteresis losses
 2. Coupling and eddy current losses
 3. Self-field losses
- Role of transport current in loss terms
- Impact of AC losses on cryogenics
- Specifying conductors based on the application

Following closely the presentation of Wilson “Superconducting magnets”

Also thanks to:

Mess, Schmueser, Wolff, “Superconducting Accelerator Magnets”

Marijn Oomen Thesis “AC Loss in Superconducting Tapes and Cables”

M.N. Wilson/ Cryogenics 48 (2008) 381–395

T. M. Mower and Y. Iwasa, Cryogenics, vol. 26, no. 5, pp. 281–292, May 1986.



Introduction



- Superconductors subjected to varying magnetic fields see multiple heat sources that can impact conductor performance and stability
- All of the energy loss terms can be understood as emanating from the voltage induced in the conductor:
 - The hysteretic nature of magnetization in type II superconductors, i.e. flux flow combined with flux pinning, results in a net energy loss when subjected to a field cycle
 - The combination of individual superconducting filaments and a separating normal-metal matrix results in a coupling Joule loss
 - Similarly, the normal-metal stabilizer sees traditional eddy currents



Magnetization losses

- The superconductor B-H cycle defines losses associated with magnetization: the area enclosed in a loop is lost as heat

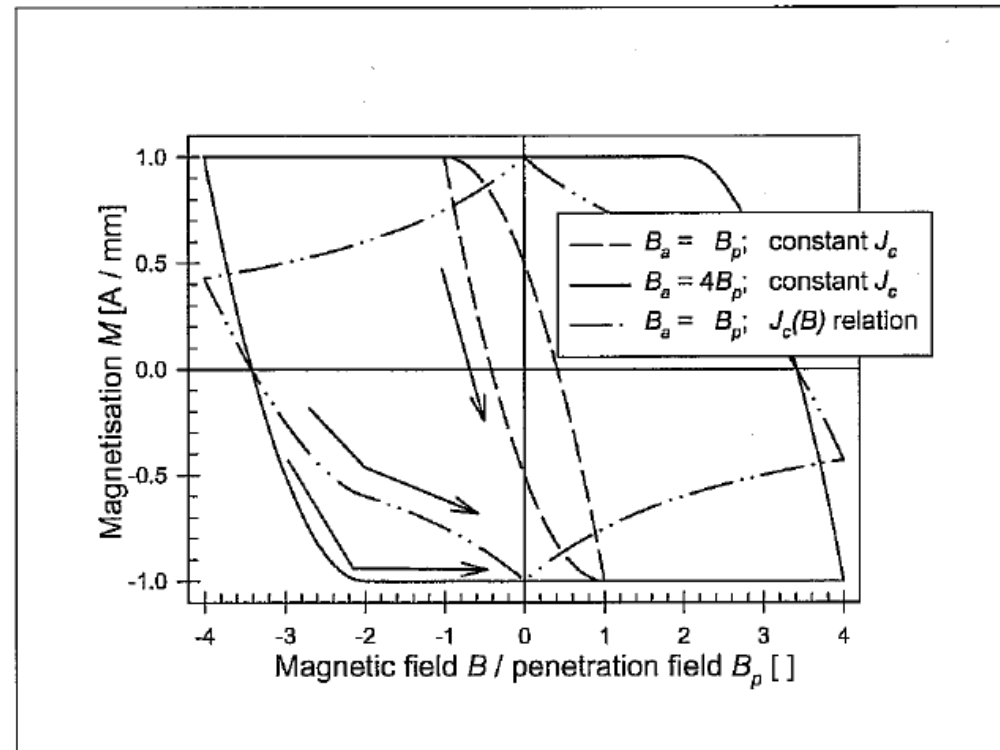
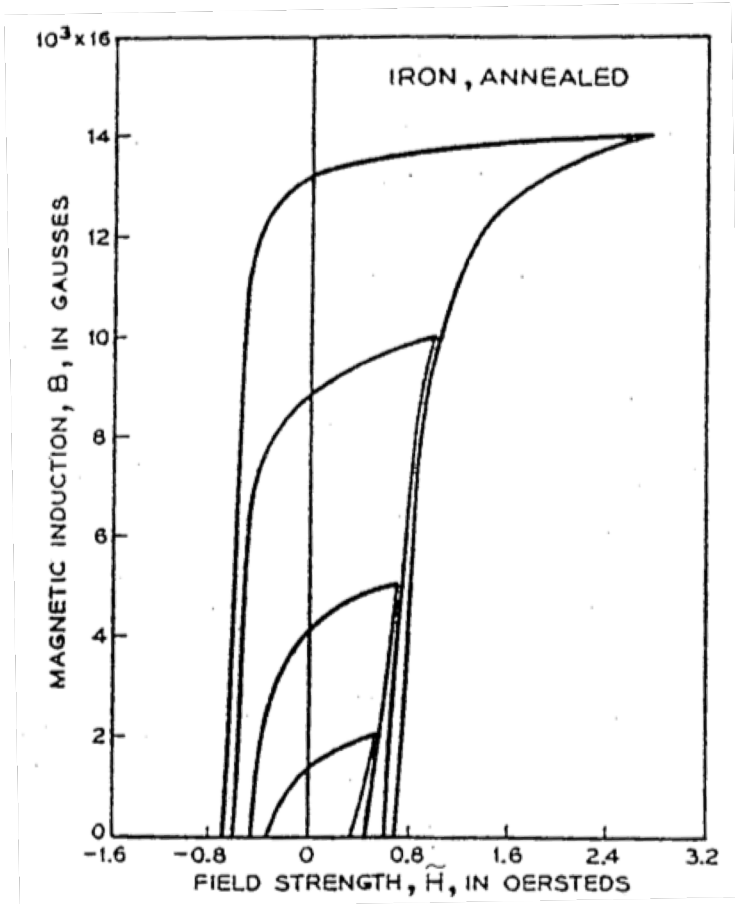


Figure 2.3 Magnetisation loops calculated for an infinite slab parallel to the magnetic field.



Hysteresis losses – basic model

Hysteresis loss is $Q = \int \vec{H} \cdot d\vec{M} = \int \vec{M} \cdot d\vec{H}$

Problem: how do we quantify this?

-Note that magnetic moment generated by a current loop I enclosing an area A is defined as

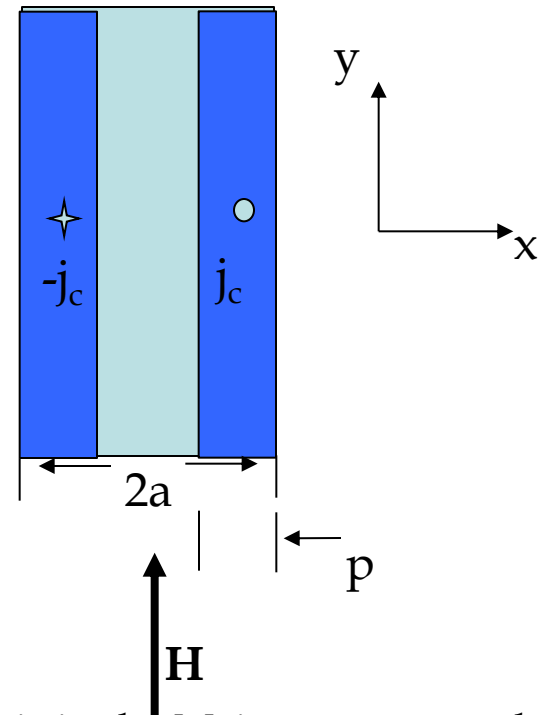
$$m = \mu_0 AI$$

$$\frac{\partial B_y}{\partial x} = \mu_0 J_c$$

The magnetization M is the sum of the magnetic moments/volume.

Assume $j=j_c$ in the region of flux penetration in the superconductor (Bean Model), then

$$\begin{aligned} \phi &= \mu_0 \int_{a-p}^a j_c x dx \\ &= \frac{\mu_0 j_c}{2} [2ap - p^2] \end{aligned}$$



- Below H_{c1} the superconductor is in the Meissner state and the magnetization from dH/dt corresponds to pure energy storage, i.e. there is no energy lost in heat;
- Beyond H_{c1} flux pinning generates hysteretic $B(H)$ behavior; the area enclosed by the $B(H)$ curve through a dB/dt cycle represents thermal loss



Calculating hysteresis losses

Some basic definitions:

B_p = Penetration field (to center)

B_m = Field modulation

$B_m = 2\mu_0 J_c \rho$ for $\rho < a$, ρ is the field penetration distance

The power generated by the penetrating field is

$$P = E_c J_c = J_c \frac{\partial \phi}{\partial t}$$

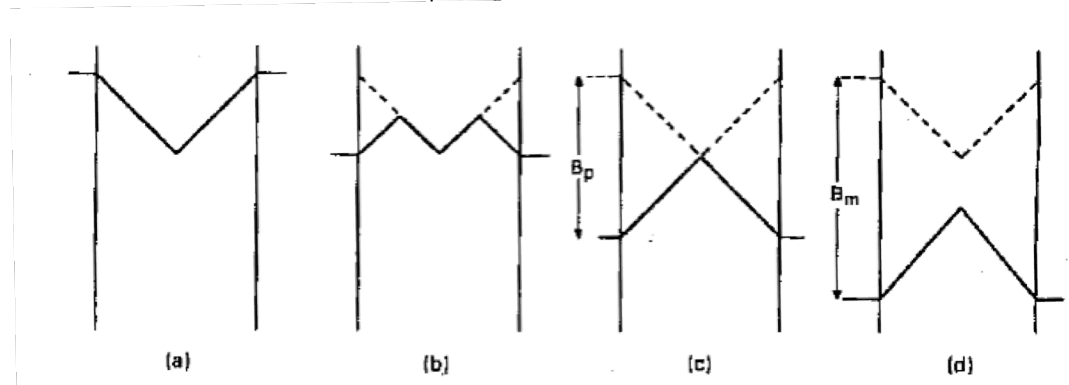
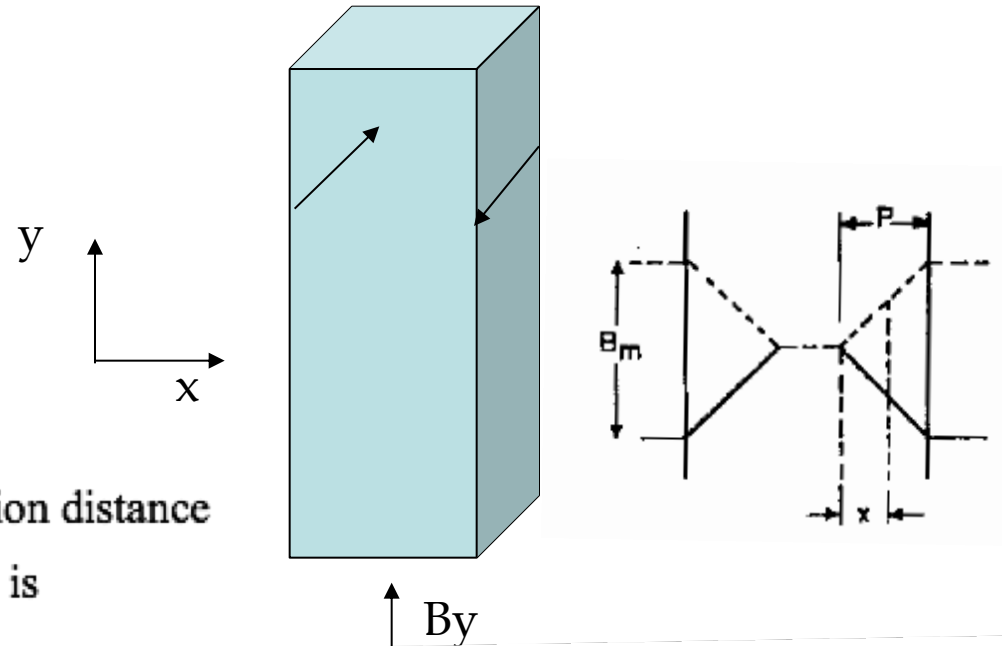


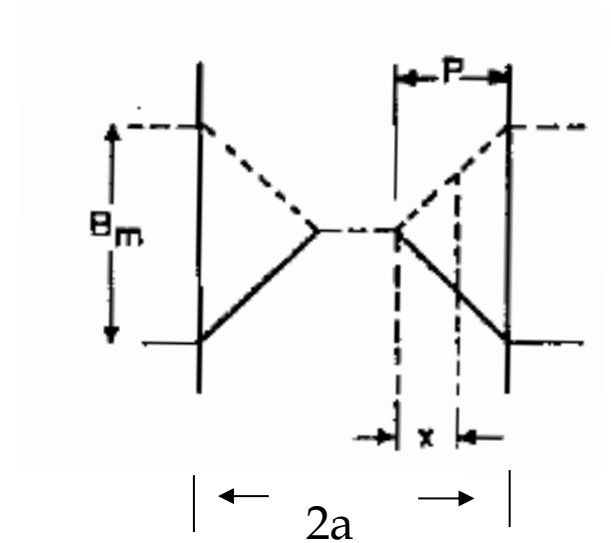
Fig. 8.4. (a) Field pattern within a superconducting slab subjected to large field change; (b) as the field is reduced; (c) when the field change penetrates to centre of slab; (d) when the field reaches a minimum value before rising again.



Calculating hysteresis losses

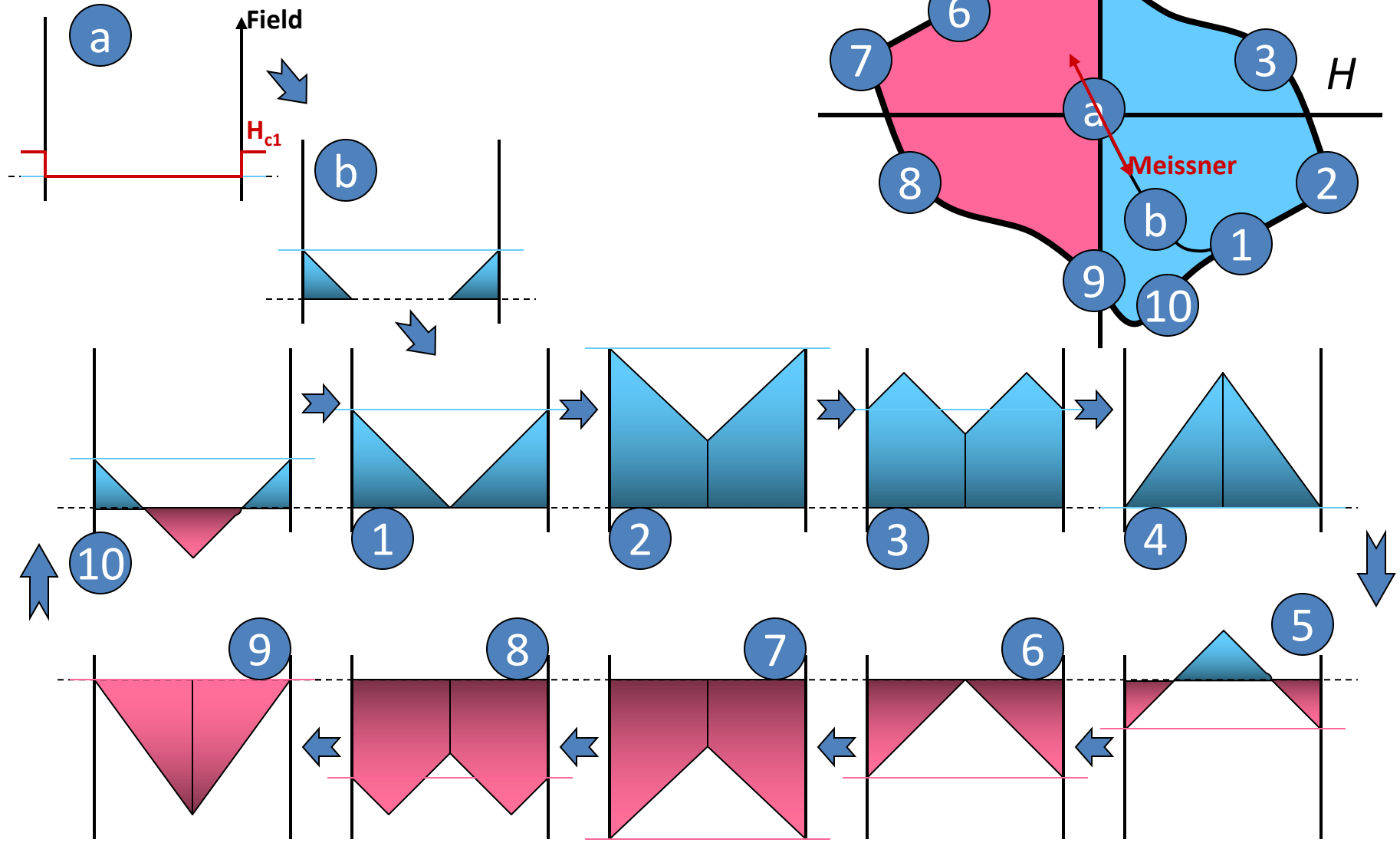
- The total heat generated for a half-cycle is then

$$\Delta\phi(x) = \int_0^x \Delta B(\xi) d\xi \approx \int_0^x \mu_0 J_c \xi d\xi = \frac{\mu_0}{2} J_c x^2$$
$$\Rightarrow q = \frac{1}{a} \int_0^p J_c (\mu_0 J_c x^2) dx = \frac{\mu_0 J_c^2 p^3}{3a}$$



- Note that this calculation assumed $p < a$; a similar analysis can be applied for the more generally case in which the sample is fully penetrated.

The Critical State





Understanding AC losses via magnetization

- The screening currents are bound currents that correspond to sample magnetization.

- Integration of the hysteresis loop quantifies the energy loss per cycle

=> Will result in the same loss as calculated using $E \circ J_c$

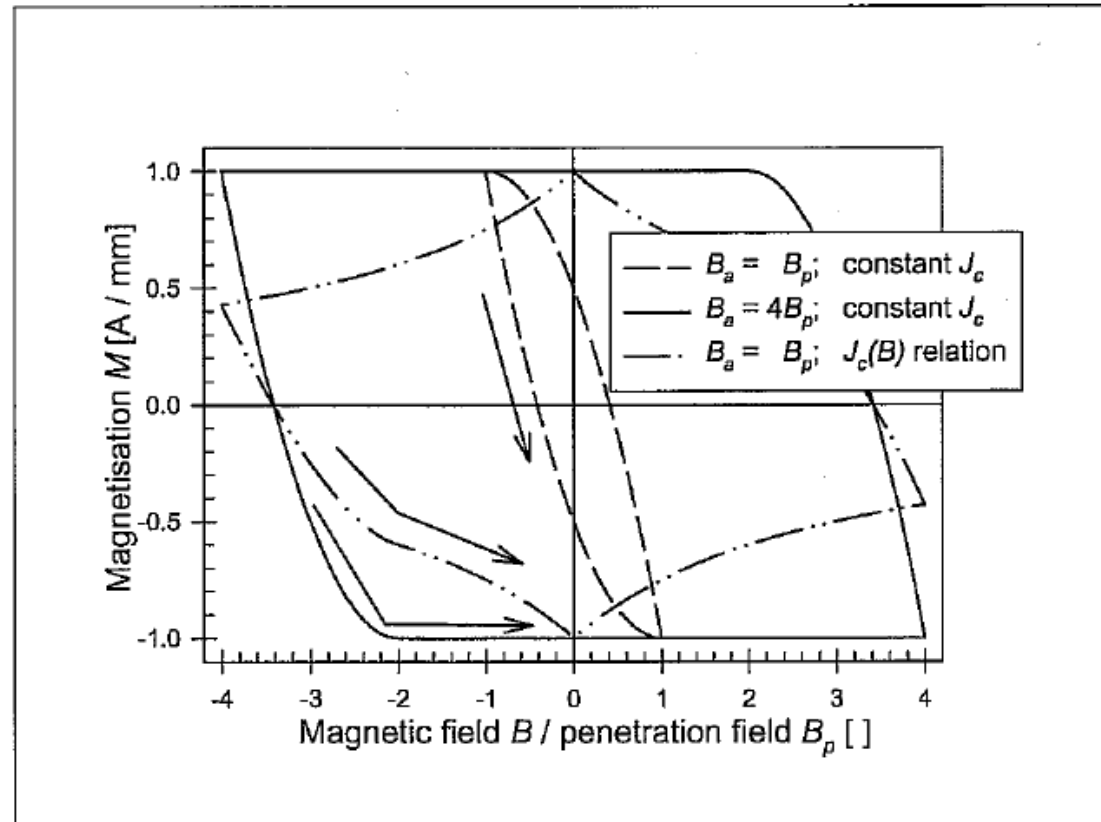


Figure 2.3 Magnetisation loops calculated for an infinite slab parallel to the magnetic field.



Hysteresis losses - general

- The hysteresis model can be developed in terms of:

$$\beta = \frac{B_m}{B_p} = \frac{B_m}{2a\mu_0 J_c}$$

The total cycle loss (for the whole slab) is then:

$$Q = \frac{B_m^2}{2\mu_0} \Gamma(\beta); \text{ The function } \Gamma \text{ (geometry dependent) has a maximum near 1.}$$

To reduce losses, we want
 $\beta \ll 1$ (little field penetration, so loss/volume is small) or
 $\beta \gg 1$ (full flux penetration, but little overall flux movement)

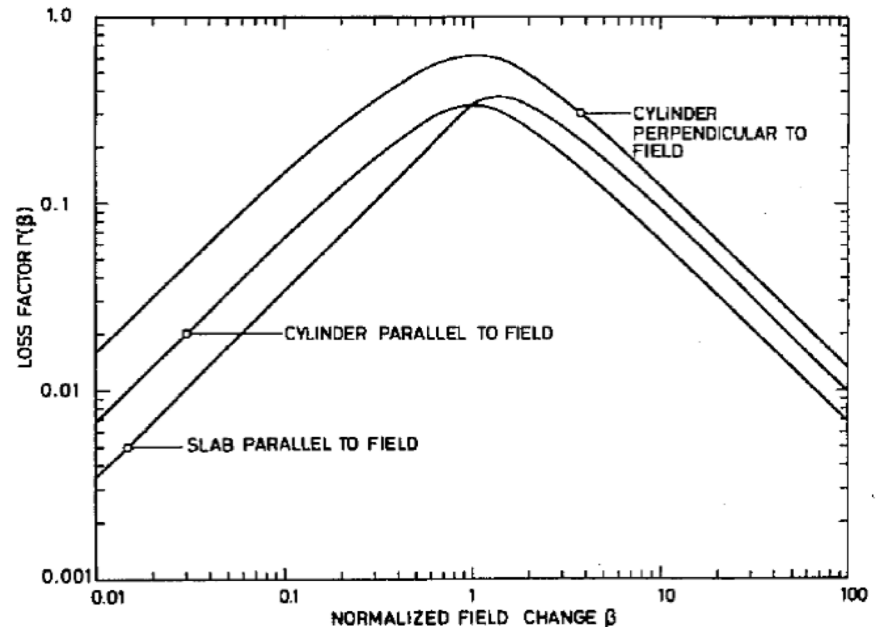
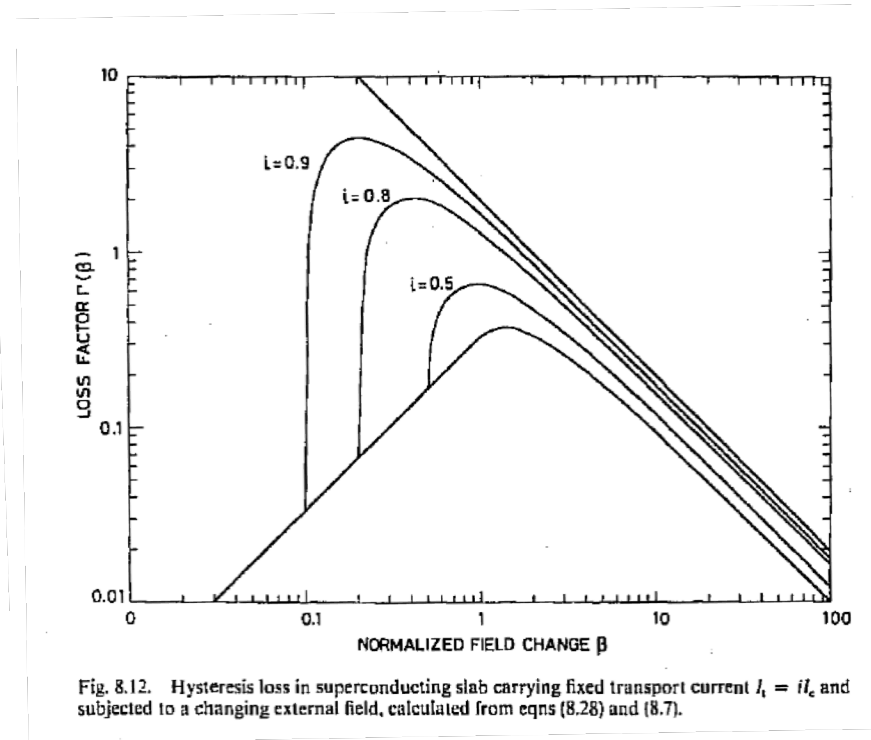
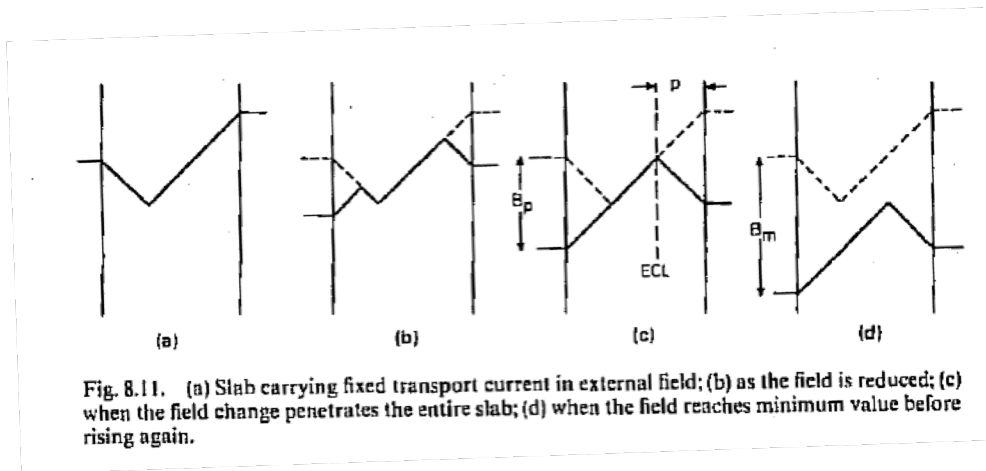


Fig. 8.5. Loss factor $\Gamma(\beta)$ for hysteresis loss per cycle in different shapes of superconductor.



Hysteresis losses

- The addition of transport current enhances the losses; this can be viewed as stemming from power supply voltage compensating the system inductance voltage generated by the varying background field.





Coupling losses

- A multifilamentary wire subjected to a transverse varying field will see an electric field generated between filaments of amplitude:

$$E = \frac{\dot{B}L}{2\pi}; \quad L \text{ is the twist-pitch of the filaments}$$

The metal matrix then sees a current (parallel to the applied field) of amplitude:

$$J = \frac{\dot{B}L}{2\pi\rho_t}$$

Similarly, the filaments couple via the periphery to yield a current:

$$J_p(\theta) = \frac{\dot{B}L \cos(\theta)}{2\pi\rho_m}$$

There are also eddy currents of amplitude:

$$J_e(\theta) = \frac{\dot{B}a \cos(\theta)}{\rho_m}$$

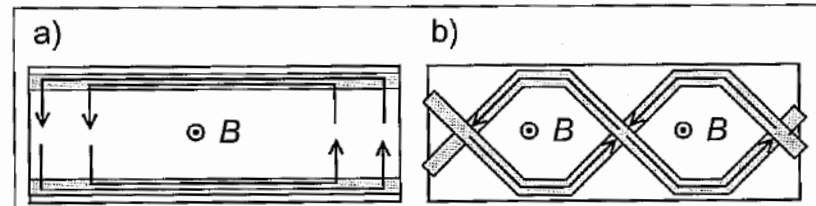
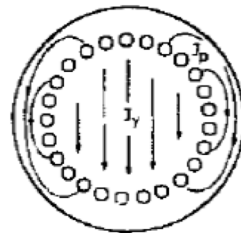
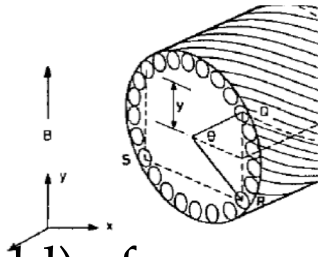


Figure 2.4 Schematic of coupling currents between two filaments in a wire or tape.



Coupling losses – time constant

- The combined $\cos(\theta)$ coupling current distribution leads to a natural time constant (coupling time constant):

$$\tau = \frac{\mu_0}{2\rho_{eff}} \left(\frac{L}{2\pi} \right)^2$$

- The time constant τ corresponds to the natural decay time of the eddy currents when the varying field becomes stationary.
- The losses associated with these currents (per unit volume) are:

$$Q_e = \frac{B_m^2}{2\mu_0} \frac{8\tau}{T_m}, \text{ where } T_m \text{ is the half-time of a full cycle}$$

Here B_m is the maximum field during the cycle.



Coupling losses – Rutherford cables

- Coupling currents also form between strands in cables

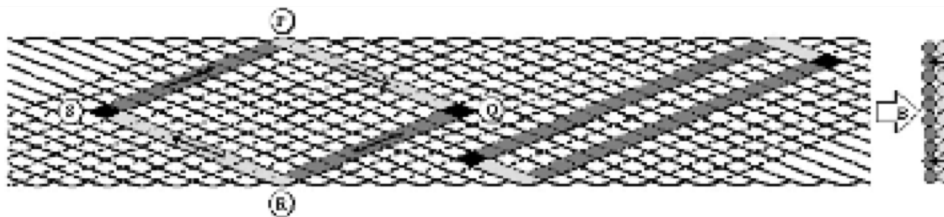


Fig. 19. Coupling currents flowing via crossover resistance R_c in transverse field (upper wires shown light grey).

$$\frac{Q_{tc}}{Q_{ia}} = \frac{R_c}{R_a} \frac{N(N-1)}{20}$$

Add core to dramatically reduce transverse coupling, while maintaining decent R_a for current sharing

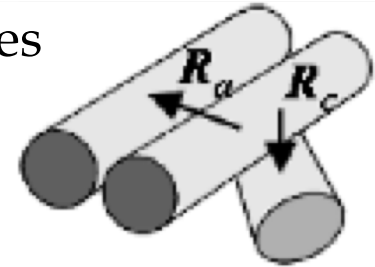
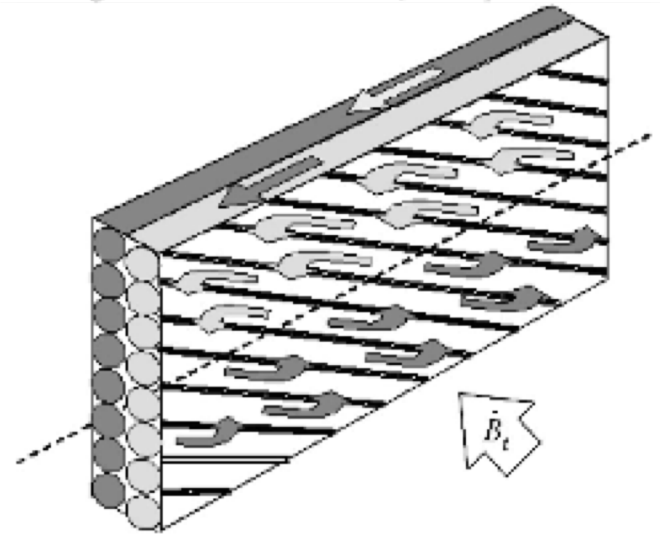
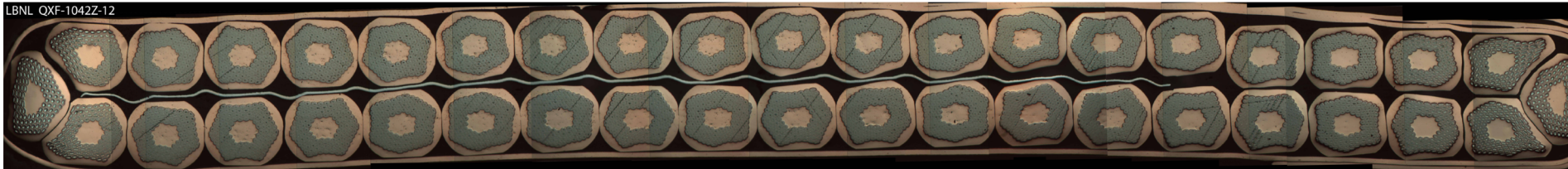


Fig. 18. Crossover resistance R_c and adjacent resistance R_a .



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Other loss terms

- In the previous analysis, we assumed the $\cos(\theta)$ longitudinal current flowed on the outer filament shell of the conductor. Depending on dB/dt , ρ , and L , the outer filaments may saturate (i.e. reach J_c), resulting in a larger zone of field penetration. The field penetration results in an additional loss term:

$$Q_p = \frac{B_m^2}{2\mu_0} \frac{4\tau^2}{T_m^2} \Gamma(\beta')$$

$$\beta' = \frac{\pi B_m}{2\mu_0 \lambda J_c a} \frac{\tau}{T_m}$$

- Self-field losses: as the transport current is varied, the self-field lines change, penetrating and exiting the conductor surface. The effect is independent of frequency, yielding a hysteresis-like energy loss:

$$Q_{sf} = \frac{B_{ms}^2}{2\mu_0} \Gamma(\beta); \quad \beta = \frac{B_{ms}}{B_p} = \frac{I}{I_c}$$

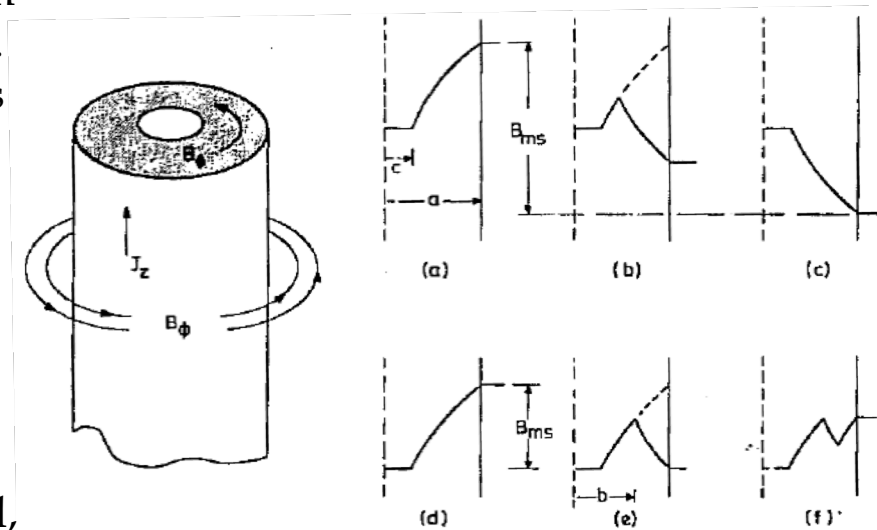


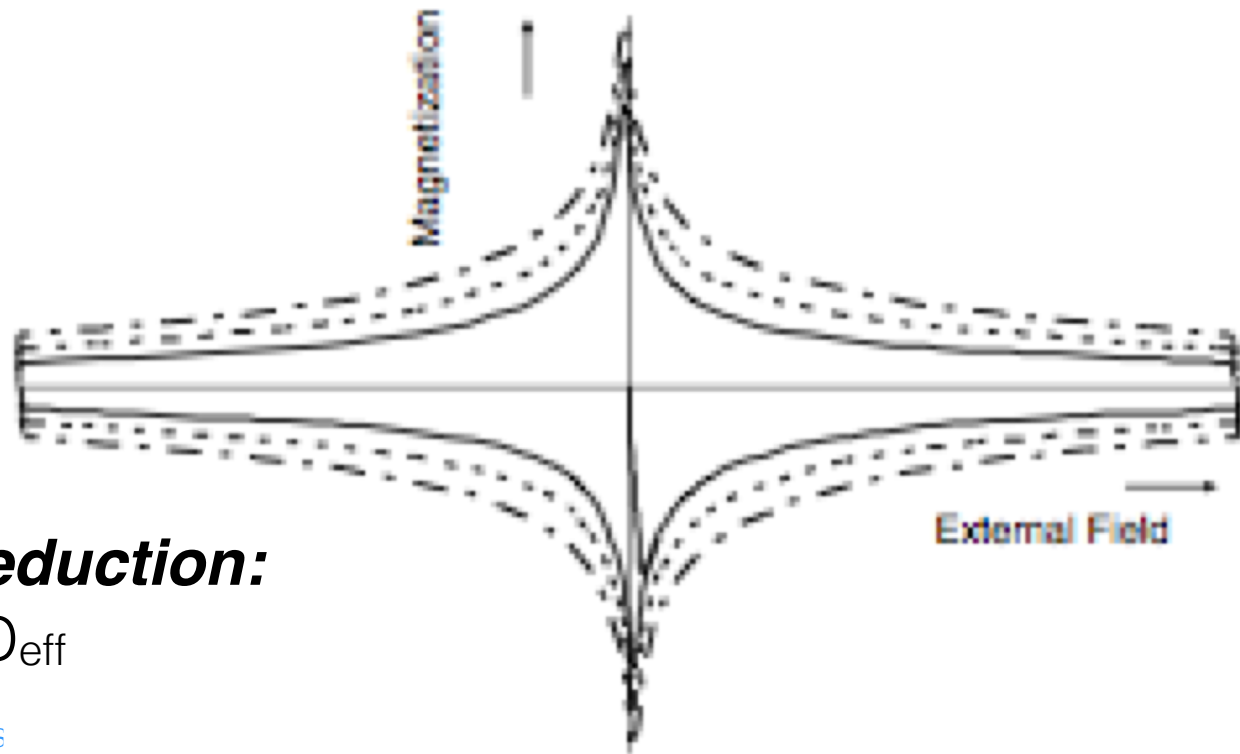
Fig. 8.24. Self-field in a superconducting cylinder or filamentary composite carrying transport current. (a), (b), and (c) show profiles of \mathbf{B} within the cylinder when transport current is reversed; (d), (e), and (f) show effect of unidirectional current oscillations.

First estimate of AC losses: Hysteresis losses

$$Q_{cyc} = \int_0^{t_0} J_c(B) \frac{2D_{eff}}{3\pi} \frac{dB}{dt} dt \quad [\text{J/m}^3, \text{ per cycle}]$$

$$Q_{hyst-tot} = Q_{cyc} * V_{sc} \quad [\text{J, per cycle}]$$

This has motivated the quest for fine filament wire!



Hysteresis loss reduction:

- minimize D_{eff}

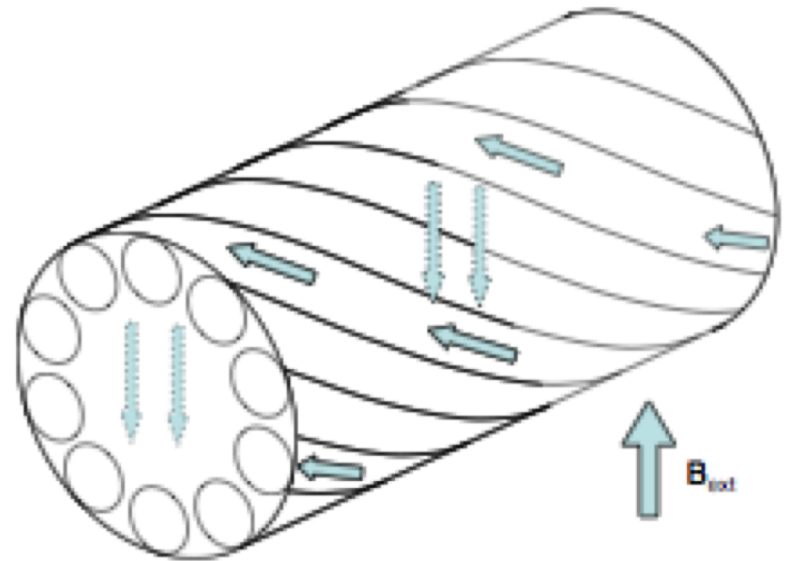
First estimate of AC losses: Coupling losses

$$\tau = \frac{\mu_0}{2\rho_t} \left(\frac{p}{2\pi} \right)^2 \quad Q_{\text{coupling}} = \frac{(dB/dt)^2}{\mu_0} 2\tau \quad [\text{W/m}^3]$$

$$Q_{\text{coupling-tot}} = Q_{\text{coupling}} * V_{\text{cond}}$$

Coupling loss reduction:

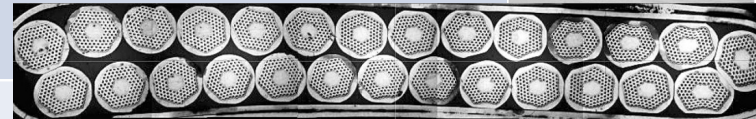
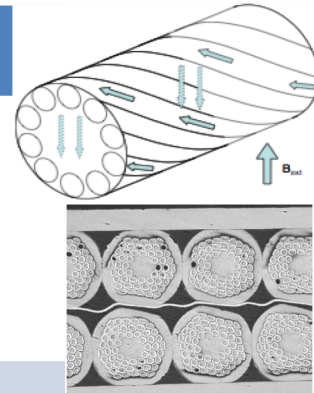
- minimize twist pitch





AC losses impact conductor, magnet and cryostat design

Loss term	Scaling	Notes
Conductor hysteresis	$P_{hys} \propto M \frac{dB}{dt} \sim J_c \frac{dB}{dt} d_{eff}$	$1 + \left(\frac{I_{max}}{I_c} \right)^2$ $\tau = \left(\frac{p}{2\pi} \right)^2 \frac{\mu_0}{2\rho_t}$
Conductor filament coupling	$P_{coup} \propto \frac{2\tau}{\mu_0} \left(\frac{dB}{dt} \right)^2$	
Cable strand coupling	$P_{cable} \propto \left(\frac{dB}{dt} \right)^2 \frac{p}{R_a} \frac{w}{t}$	
Eddy currents		



- **Loss estimates are further complicated by field regime, operational current, etc.**
- **Final design is a balance between heat capacity, losses, heat transfer and duty cycle resulting in conductor temperature excursions and hence performance limitations**

- **Hysteresis:**
 - Reduce d_{eff}
 - Increases with I/I_c
- **Coupling:**
 - Minimize twist pitch
 - Modify inter-filament resistance
- **Eddy currents:**
 - laminations



Use of the AC-loss models



- It is common (but not necessarily correct) to add the different AC loss terms together to determine the loss budget for an conductor design and operational mode.
- AC loss calculations are “imperfect”:
 - Uncertainties in effective resistivities (e.g. matrix resistivity may vary locally, e.g. based on alloy properties associated with fabrication; contact resistances between metals may vary, etc)
 - Calculations invariably assume “ideal” behavior, e.g. Bean model, homogeneous external field, etc.
- For real applications, these models usually suffice to provide grounds for conductor specifications and/or cryogenic budgeting
 - For critical applications, AC-loss measurements (non-trivial!) should be undertaken to quantify key parameters



Special cases: HTS tapes

- HTS tapes have anisotropic J_c properties that impact AC losses.
- The same general AC loss analysis techniques apply, but typical operating conditions impact AC loss conclusions:
 - the increased specific heat at higher temperatures has significant ramifications - enhances stability
 - Cryogenic heat extraction increases with temperature, so higher losses may be tolerated

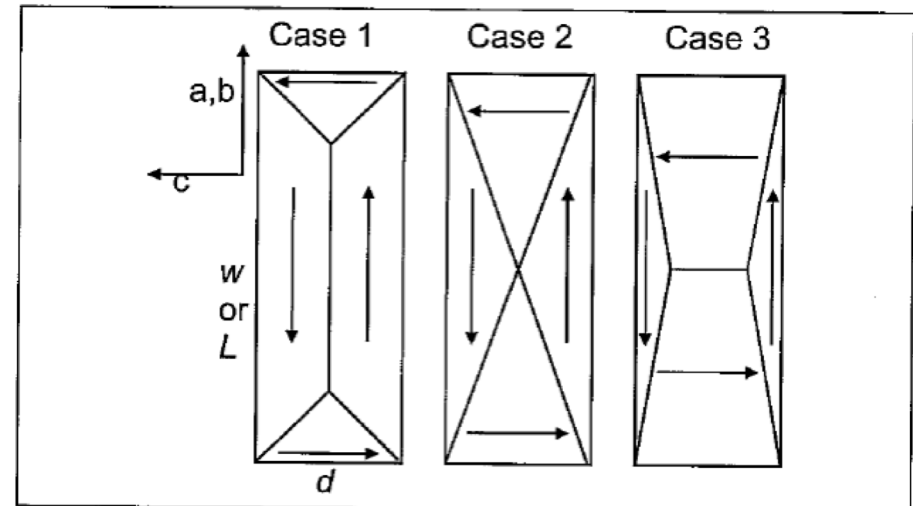


Figure 2.11 Screening currents in a slab with anisotropic critical-current density.



AC losses and cryogenics



- The AC loss budget must be accounted for in the cryogenic system
 - Design must account for thermal gradients – e.g. from strand to cable, through insulation, etc. and provide sufficient temperature margin for operation
 - Typically the temperature margin needed will also depend on the cycle frequency; the ratios of the characteristic cycle time (τ_w) and characteristic diffusion time (τ_d) separates two regimes:
 1. $\tau_w \ll \tau_d$: Margin determined by single cycle enthalpy
 2. $\tau_w \gg \tau_d$: Margin determined by thermal gradients
- The AC loss budget is critical for applications requiring controlled current rundown; if the AC losses are too large, the system may quench and the user loses control of the decay rate



Specifying conductors for AC losses



- As a designer, you have some control over the ac losses:
 - Control by conductor specification
 - Filament size
 - Contact resistances
 - Twist pitch
 - Sufficient temperature margin (e.g. material T_c , fraction of critical current, etc)
 - Control by cryogenics/cooling
 - Appropriate selection of materials for good thermal conductivity
 - Localization of cryogen near thermal loads to minimize ΔT
- Remember: loss calculations are imperfect! For critical applications, AC loss measurements may be required, and some margin provided in the thermal design to accommodate uncertainties