

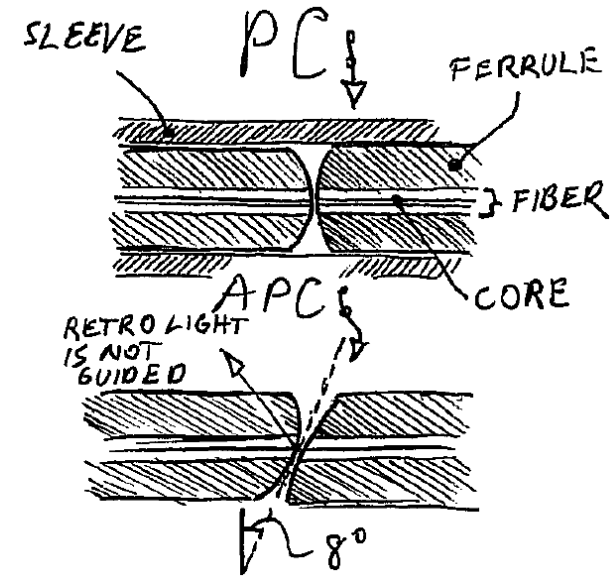
Part 4: fiber components

Fiber optic system components

- **Connectors**
- **Directional couplers**
- **Lens couplers**
- **Attenuators**
- **Delays**
- **Polarization controllers**
- **Isolators**
- **Filters**
- **Modulators**
- **Amplifiers**
- **Detectors**

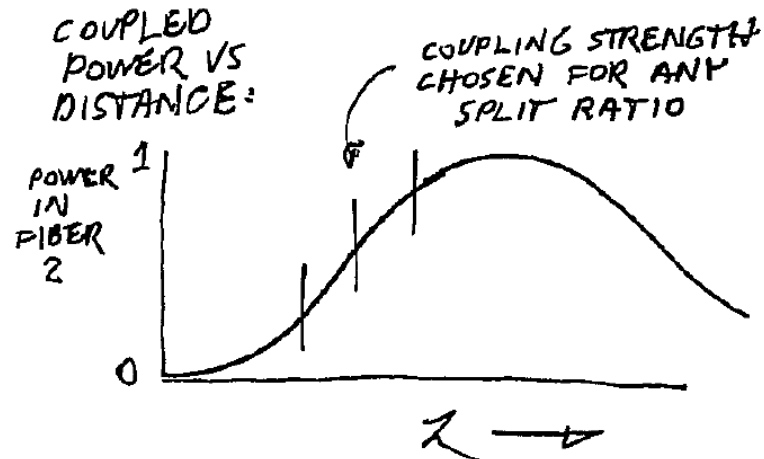
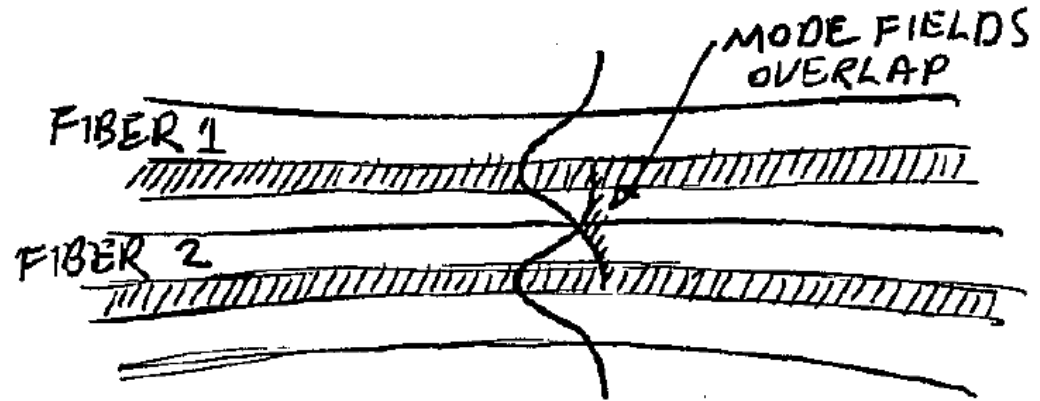
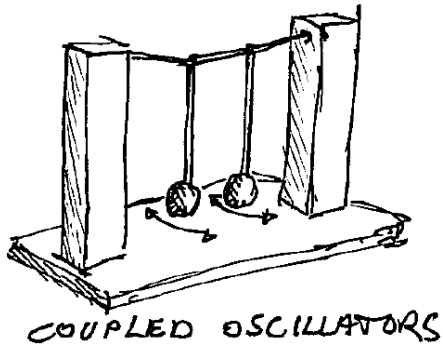
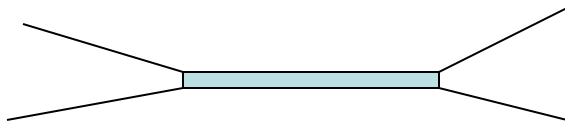
Connectors

- **Cores must be aligned to sub-micron accuracy**
 - **Sub-micron concentricity between core, cladding, ferrule, sleeve**
 - **Cores in contact avoid loss at interface (PC)**
 - **Loss through is ~0.2dB average, though variable and non-repeatable**
 - **Ends are polished convex, to ensure core contact**
 - **Still some back reflection at discontinuity (see RWV p. 292, esp. 294)**
- **Some back reflection still exists, minimized by 8 degree angle (APC)**
 - **Back reflection loss >60dB**
- **Screw-on and push-on styles (FC, SC, LC)**
- **Mechanical splices**
 - **Push flat cleaved fibers together in a tube or groove**
 - **Fibers are cleaved flat using a good quality "cleaver"**
 - **High loss (~2 dB), but quick and cheap**
- **Fusion splices**
 - **Melt fibers together with hot arc or filament**
 - **Lowest loss (<0.02dB), negligible back reflection**
 - **Align cladding, assume concentricity, or align cores by passing light through**
 - **Some skill, expensive machine needed (\$5-100k, depending on complexity)**



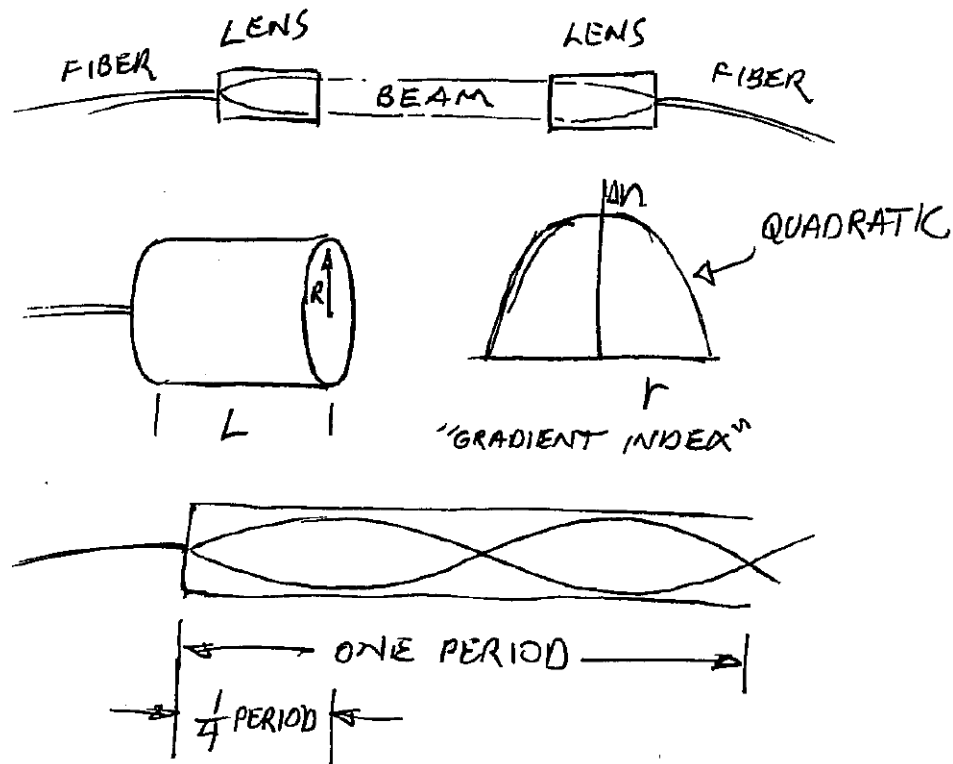
Directional couplers

- Evanescent field in cladding couples to nearby core
- High directivity, low loss, easily manufacturable in any coupling ratio
- Tree for multiple splits



Lens coupler

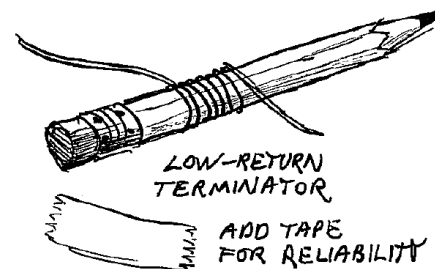
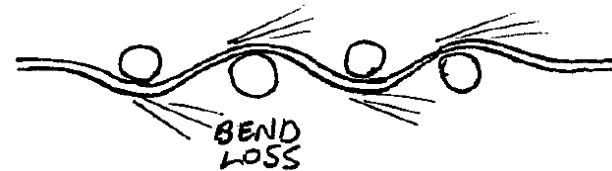
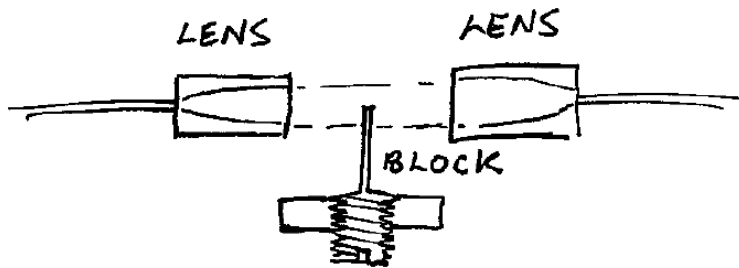
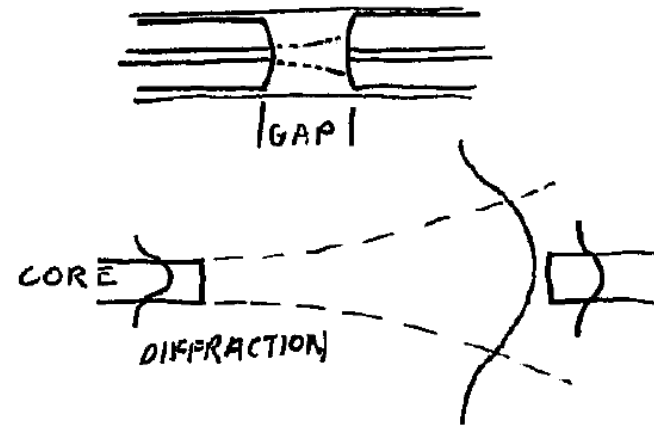
- To break out of fiber for a short distance, to insert some optical element
- Use lenses to make collimated beam between fibers
- Typical gradient index (GRIN) lens (RWV page 754)
- Less than 1dB loss



Attenuators

- **Pulling connectors apart (see RWV p. 783)**
 - Adjustable
 - Fixed ring
- **Blocking free space beam**
 - Digitally controlled versions
- **Bend loss**
 - Loops or s-curves
- **Short sections of lossy fiber**
 - Packaged like RF attenuators

GAP ATTENUATOR:

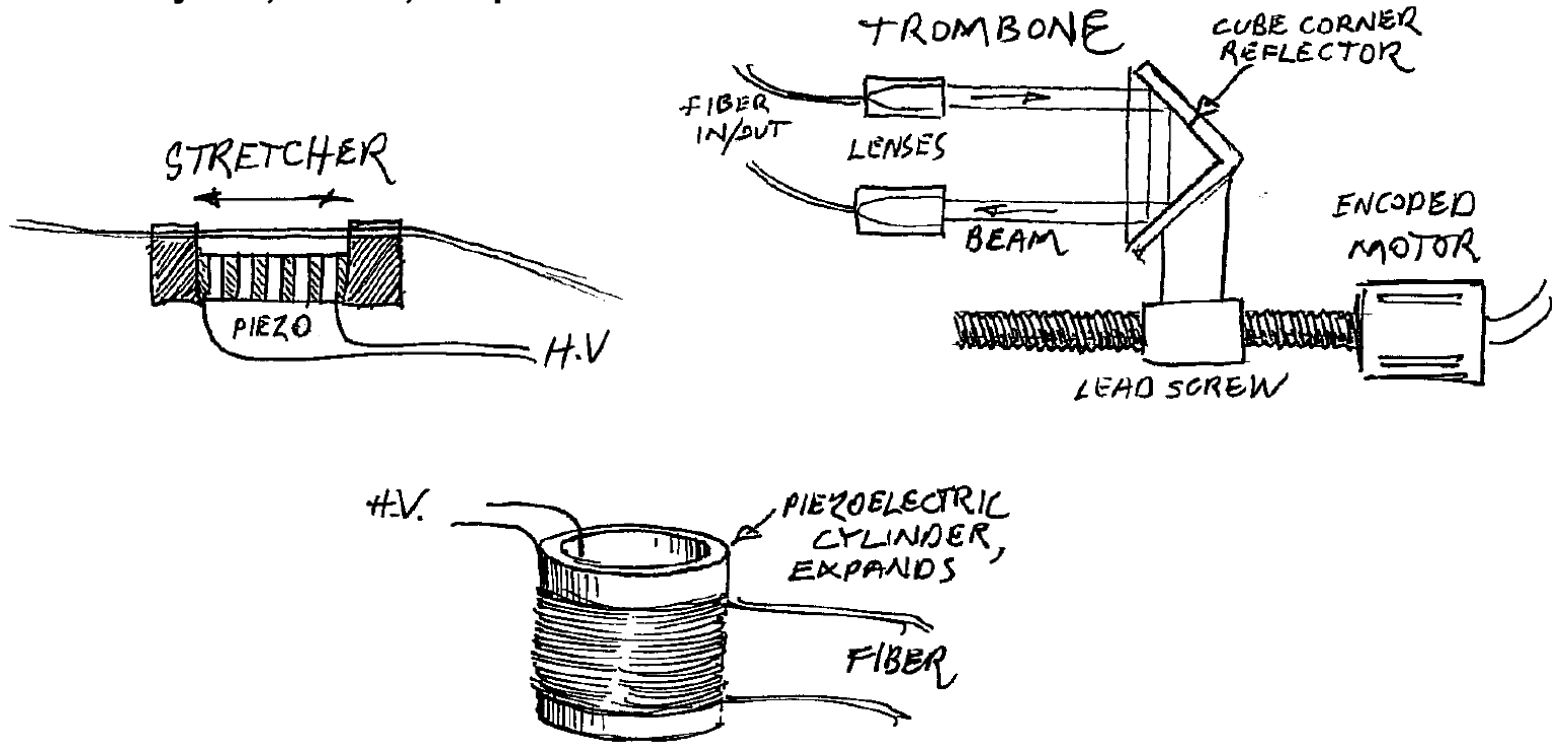


Variable phase/time delays

- **Optical trombone**
- **Piezo fiber stretcher**
- **Heated fiber spool**
- **AO fiber frequency shifter**
- **Optical IQ modulator**
- **RF IQ modulator**

Trombone, stretcher, heater

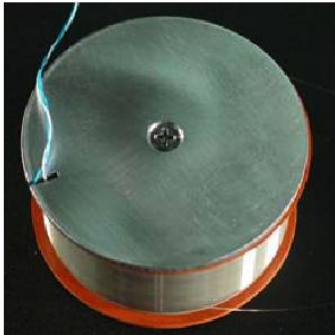
- **Optical trombone**
 - A moving mirror or prism on a rail with motor
 - Slow, long range, few dB loss, stable with power removed
- **Piezo fiber stretcher**
 - Pulls on fiber to change delay
 - Equation, note stress and lengthening both
 - Acoustic frequency, low loss, short range ($\sim 10^{-4}$ to 10^{-5} of total length, 12ps for 40m spool)
- **Heated fiber spool**
 - $\sim 10^{-5}/C$ thermal delay coefficient
 - Very slow, low loss, cheap



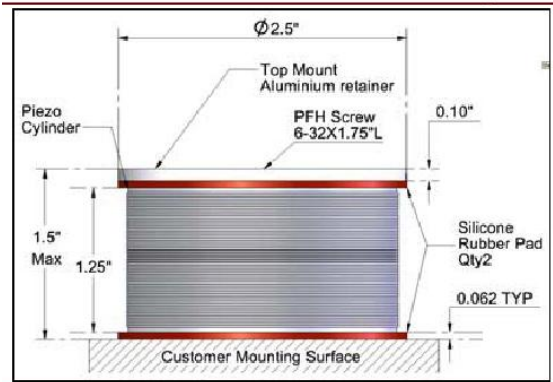
Piezo stretcher

$$\Delta t/t = \epsilon \left\{ 1 + -\frac{n_o^2}{2} [P_{12} - v(P_{11} + P_{12})] \right\}$$

$$= (0.78 \text{ to } 0.81)\epsilon \text{ (depending on ref.)}$$



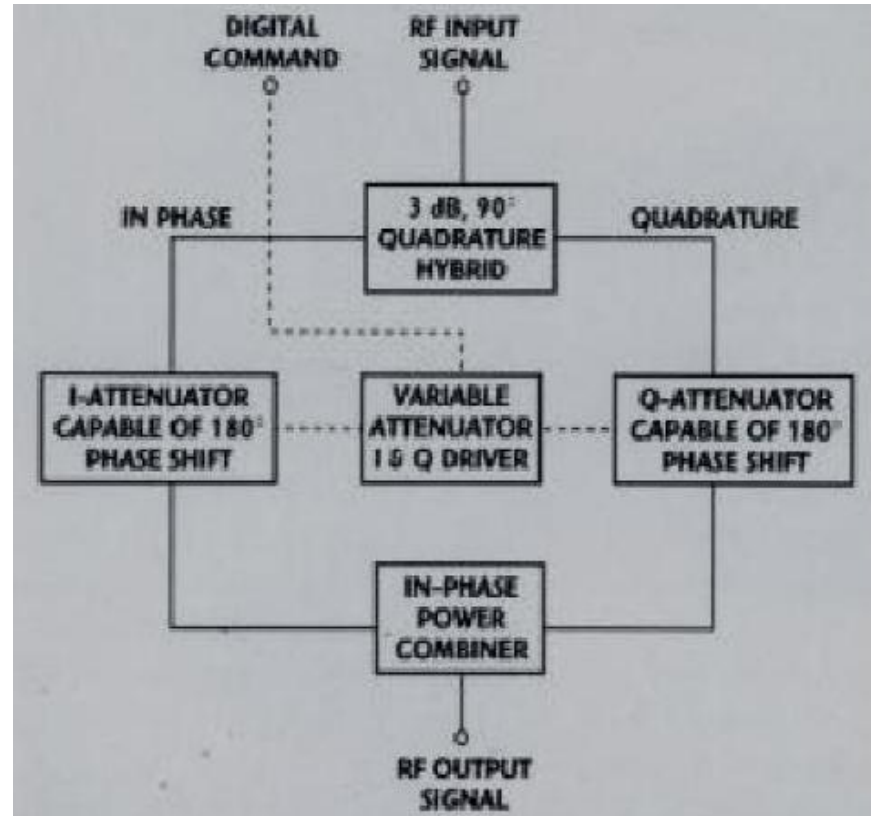
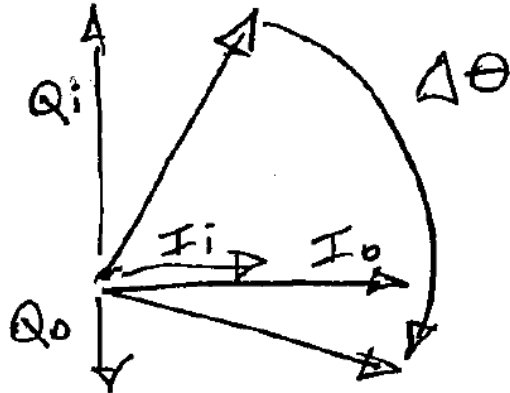
Bare Lead Fiber Stretcher with Mounting Kit



MODULATOR	SMF-28 FIBER
Operational Wavelengths	1260 to 1625 nm
Modulation Constant [low frequency]	27 radians/V @ 1.3 μm 23 radians/V @ 1.55 μm
Fiber Stretch / Optical Delay	3.8 $\mu\text{m}/\text{V}$, 0.028 ps/V
Linearity	3% at full scale drive
Frequency Range	See chart page 2
Optical Loss	≤ 0.5 dB, typical 0.2 dB
Extinction Ratio	n/a
Maximum Voltage Range [low frequency]	$\pm 400\text{V}$ [800V P-P]
Impedance [off resonance]	Capacitance 0.1 μF , floating
Wire Lead	18 inches, flying leads, #30
Operational Temperature Range	0° to 70° C
Fiber Type	SMF-28e, 250 μm acrylate jacket
Fiber Length	40 meters [includes 1 m bare fiber leads]

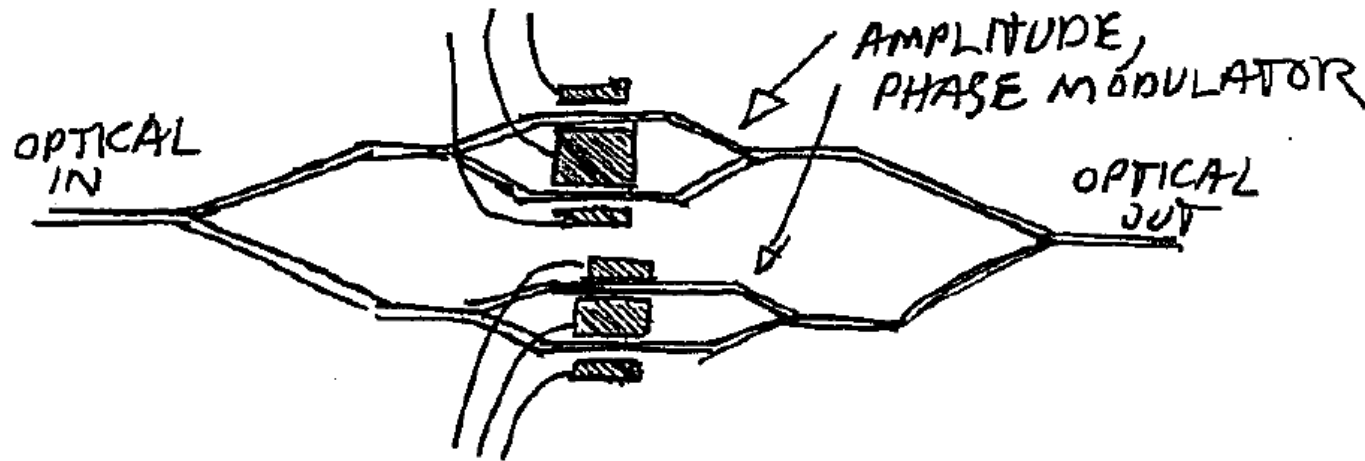
RF in-phase and quadrature (I&Q) modulator

- Input wave separated into two components, one shifted $\pi/2$ in phase
- Both components can be multiplied by [-1 to 1]
- They are added at the output to produce a phase shifted wave
 - If the control input is a continuous frequency, the modulator becomes a frequency shifter



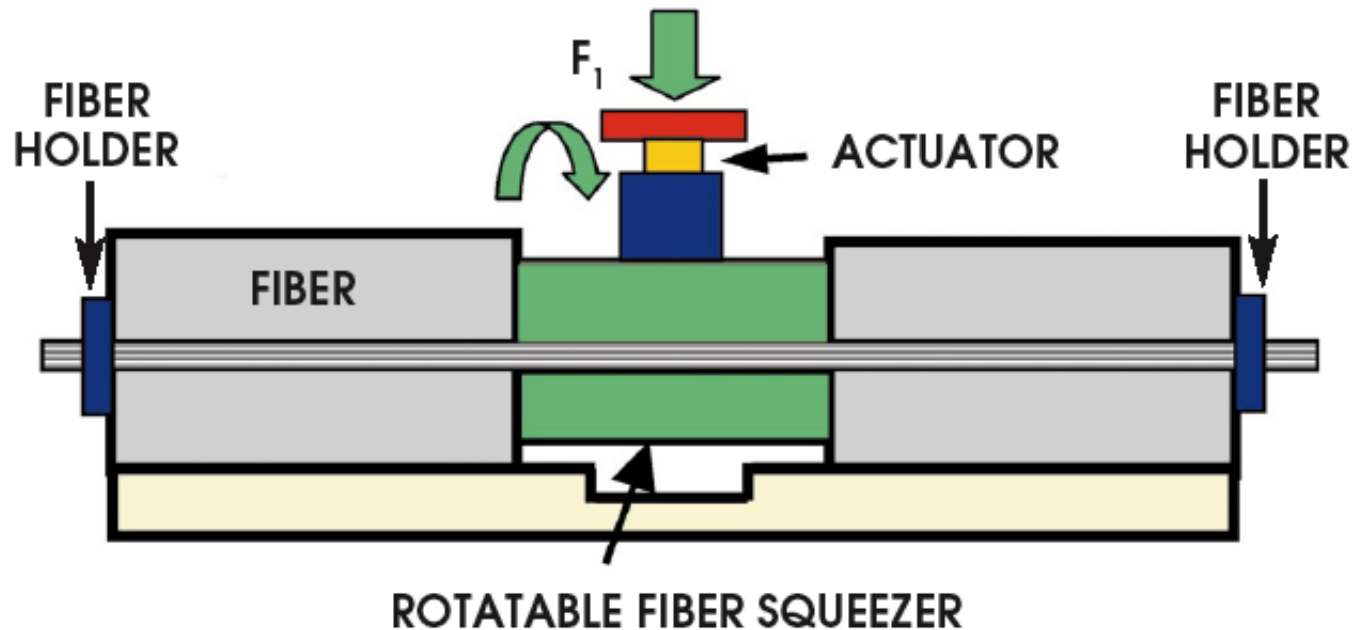
Optical I and Q modulator

- **Input signal is split equally in two arms**
- **Amplitude and phase can be controlled in both arms**
 - **Note that if both arms of a Mach-Zehnder modulator are driven together, the device becomes a pure phase modulator**
- **When recombined, the resultant output wave has controllable phase**
- **Used in Differential Phase Shift Keying (DPSK) telecommunications**
 - **Populate the complex plane with a ring of bit values (4 or 8)**



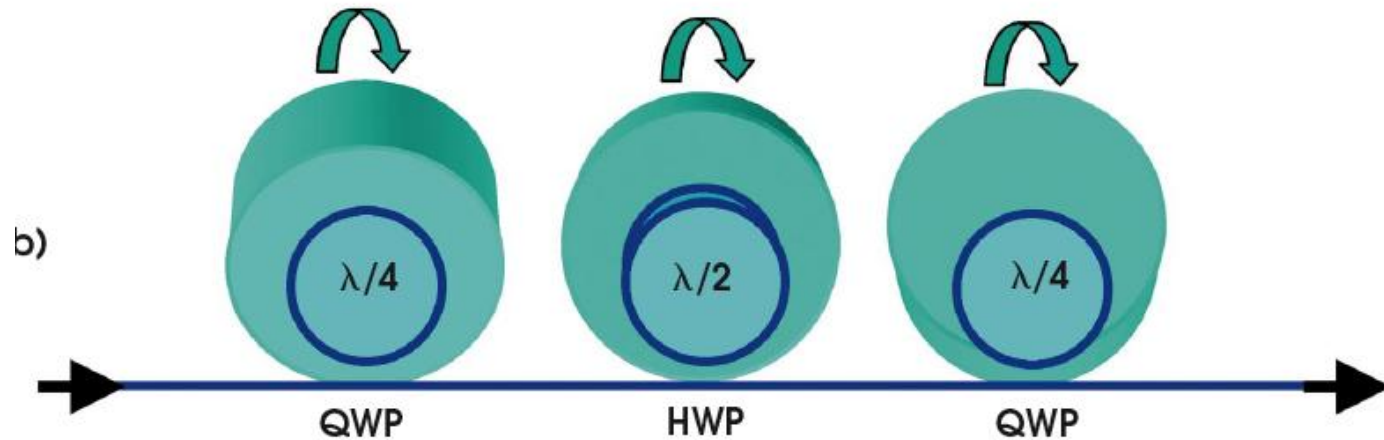
Stress birefringence polarization controller

- **Variable stress causes variable relative retardation parallel and perpendicular**
- **Rotation of the retarder with respect to the input polarization**
 - **Free space optic embodiment is Babinet compensator**



Loop polarization controller

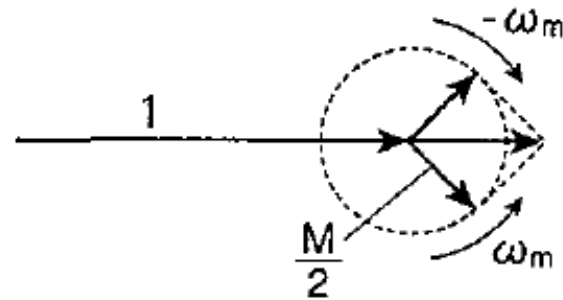
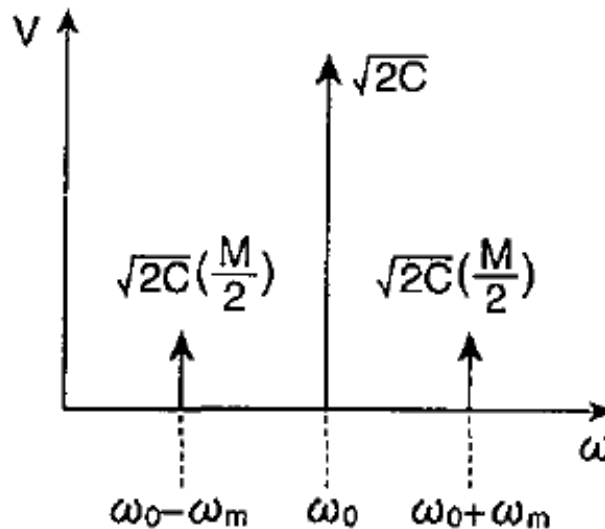
- **Looping of fibers induces stress birefringence**
- **Loops (“paddles”) are rotated with respect to the input and each other**
- **Arrangement allows access to any part of the Poincare sphere from any other part**
- **H. C. Lefevre, Electronics Letters 16, 778 (1980)**



Amplitude modulation

- **M is modulation depth**
- **Note that spectrum analyzer will detect sideband power, so squaring carrier/sideband ratios**

$$\begin{aligned} V(t) &= \sqrt{2C} [1 + M \cos \omega_m t] \sin \omega_0 t \\ &= \sqrt{2C} \left[\sin \omega_0 t + \frac{M}{2} \sin(\omega_0 + \omega_m)t + \frac{M}{2} \sin(\omega_0 - \omega_m)t \right] \end{aligned}$$



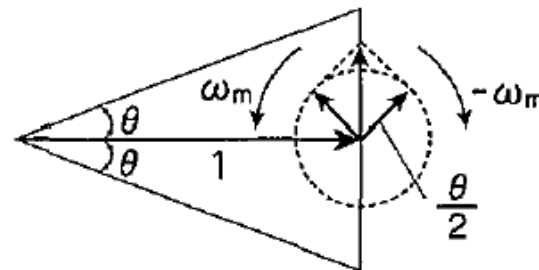
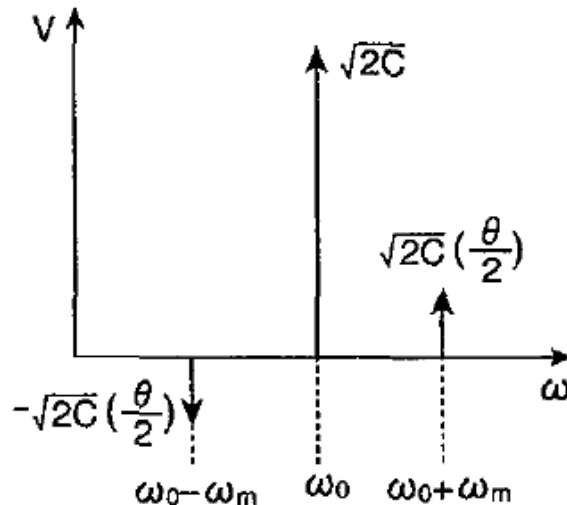
Phase modulation

$$V(t) = \sqrt{2C} \sin[\omega_0 t + \theta \sin \omega_m t],$$

- **Theta is the modulation depth, which is the peak phase deviation in radians**
- **If the “pi phase shift” voltage is V_p , the modulation depth is $\pi * V/V_p$**

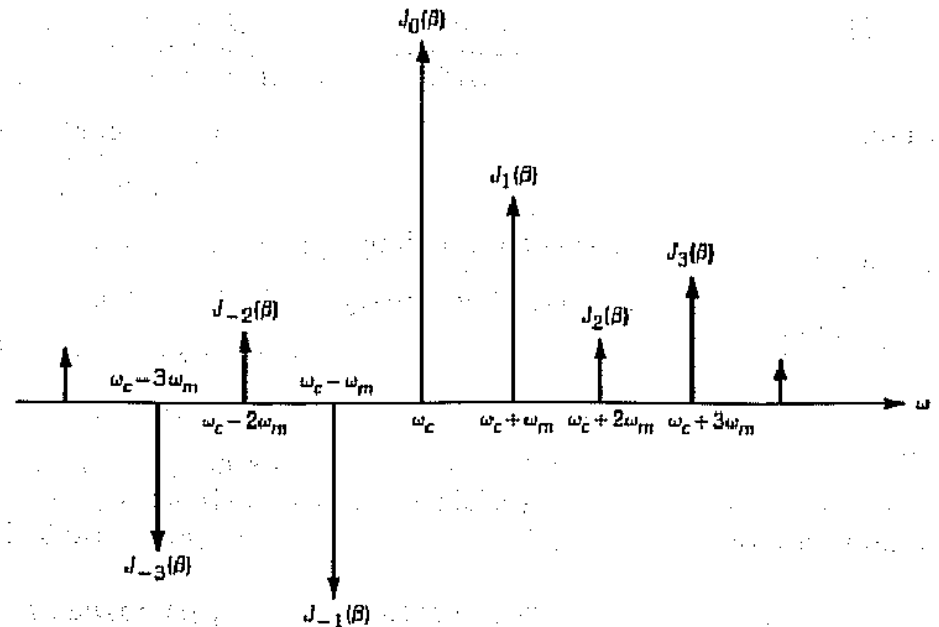
$$V(t) = \sqrt{2C} J_0(\theta) \sin \omega_0 t + \sqrt{2C} \sum_{n=1}^{\infty} J_n(\theta) \sin(\omega_0 + n\omega_m)t$$

$$+ \sqrt{2C} \sum_{n=1}^{\infty} (-1)^n J_n(\theta) \sin(\omega_0 - n\omega_m)t.$$

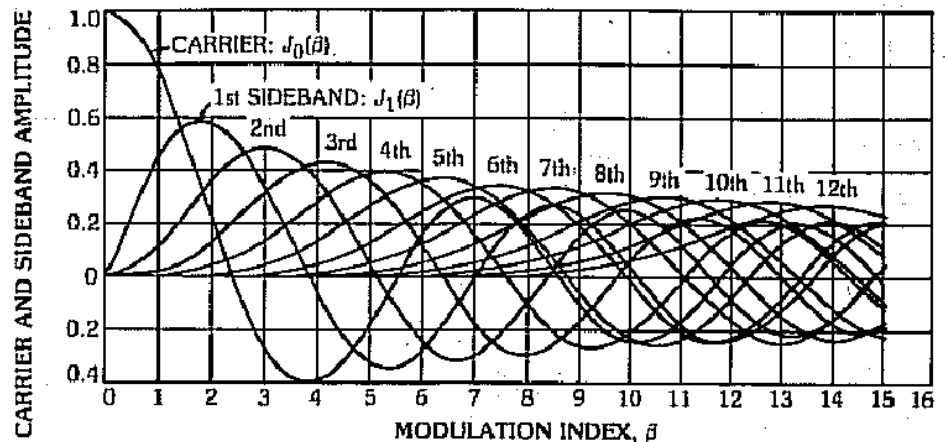


High-index phase modulation

- Sideband amplitudes are the Bessel functions
- One can observe the sideband power, the square of the amplitudes, with a tunable filter
- Mod index is easily determined by noting the relative sideband heights



$$\begin{aligned}
 e(t) &= A_c [J_0(\beta) \cos \omega_c t \\
 &\quad - J_1(\beta) [\cos(\omega_c - \omega_m)t - \cos(\omega_c + \omega_m)t] \\
 &\quad + J_2(\beta) [\cos(\omega_c - 2\omega_m)t + \cos(\omega_c + 2\omega_m)t] \\
 &\quad - J_3(\beta) [\cos(\omega_c - 3\omega_m)t - \cos(\omega_c + 3\omega_m)t] + \dots] \\
 &= A_c \sum_{n=-\infty}^{\infty} J_n(\beta) \cos(\omega_c + n\omega_m)t
 \end{aligned}$$

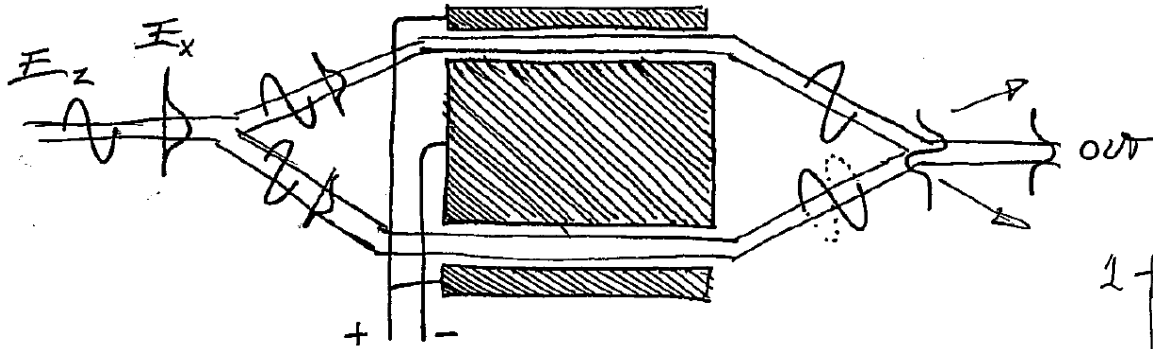


Modulators, electro-optic

- **Waveguide on electro-optic substrate (RWV p. 768)**
 - **Phase**
 - **Waveguide sort of like a fiber**
 - **Waveguide material changes index as a function of external field**
 - **Field applied over a few microns, so low voltage**
 - **Phase match RF and light for high speed (~100GHz)**
 - **Amplitude**
 - **Turns phase modulation into amplitude modulation, by putting phase modulator in an interferometer**
 - **When phase difference is π , light is extinguished because field at output looks like higher order mode, doesn't propagate**
 - **Note: not an amplitude modulator but a power modulator, thus AM sidebands are different**
- **RWV page 709, 710**

EO amplitude (power) modulator

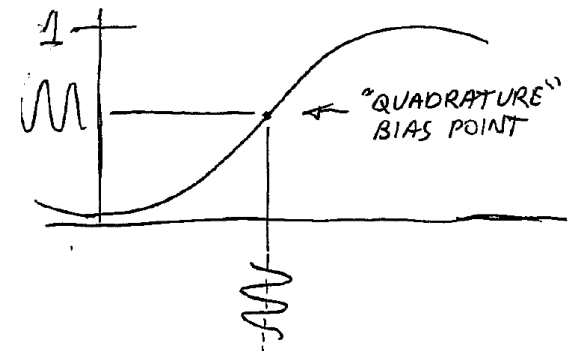
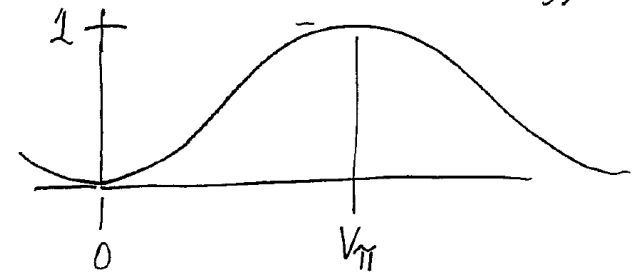
- Phase



$$P(t) = P_0 \sin^2 \left[\frac{1}{2} \left(\pi \frac{V(t)}{V_{\pi}} + \phi_B \right) \right]$$

$$= P_0 \frac{1}{2} \left(1 - \cos \left(\pi \frac{V(t)}{V_{\pi}} + \phi_B \right) \right)$$

$$\sin^2 \theta = \frac{1}{2} (1 - \cos 2\theta)$$

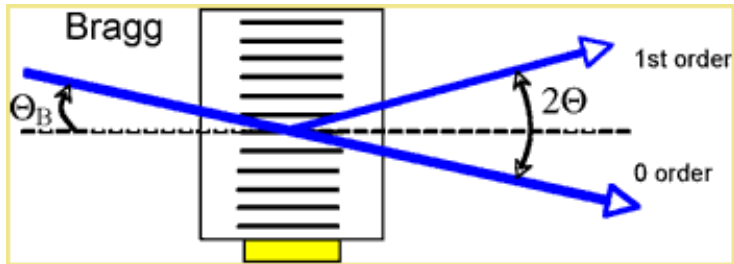


Modulators, acousto-optic

- **Bragg diffraction** interacts acoustic wave with optical wave
- **Periodic index modulation** can be created by acoustic wave (stress-optic effect), launched by RF (piezoelectric effect)
- **Index modulation** is moving at acoustic velocity, Doppler shifting the optical wave up or down depending on geometry
 - The frequency shift is the RF frequency
 - The added phase shift is the RF phase
- **Diffraction efficiency** depends on index wave amplitude and interaction length
 - **Diffraction efficiency** can be modulated by changing RF amplitude, resulting in amplitude modulator
- **Other functions** such as tunable filters and scanners can be realized using swept RF sources

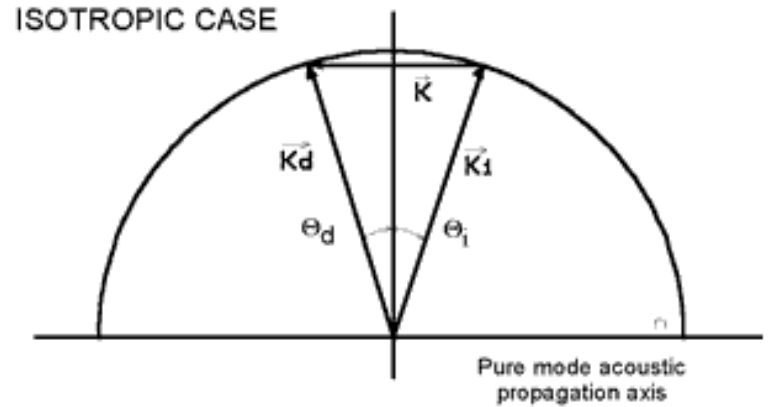
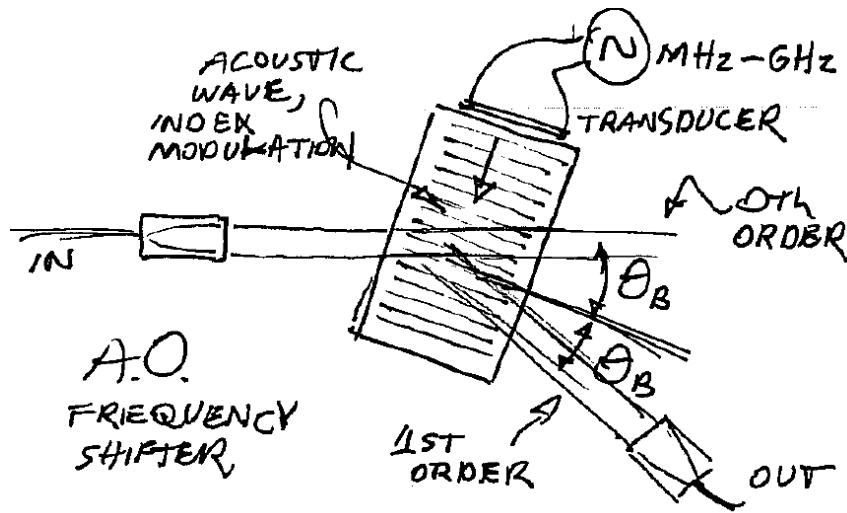
$$n(z, t) = n + \Delta n \cos(\omega t - kz),$$

$$\Delta n = -\frac{1}{3}n^3 p_{ij} a_j$$



AO frequency shifter

- Constant amplitude RF applied to modulator
- Frequency shift is RF frequency
- Efficiency of diffraction into first order ~60%
- Couple light out of fiber and back in after diffraction

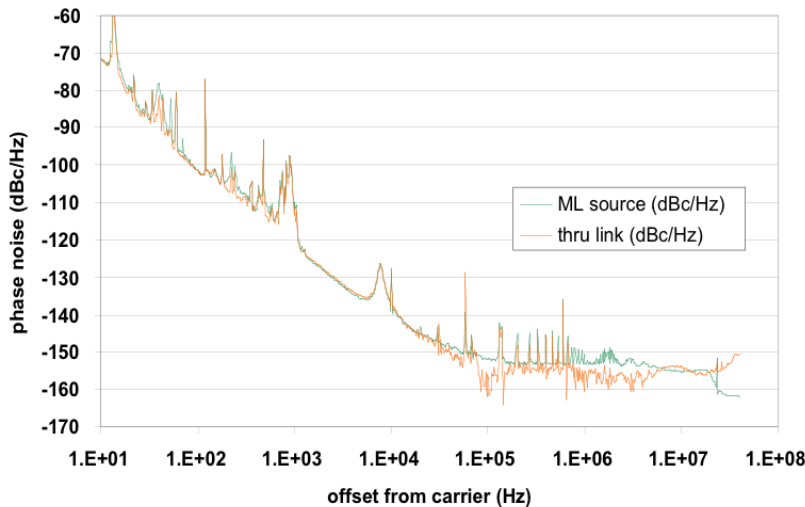
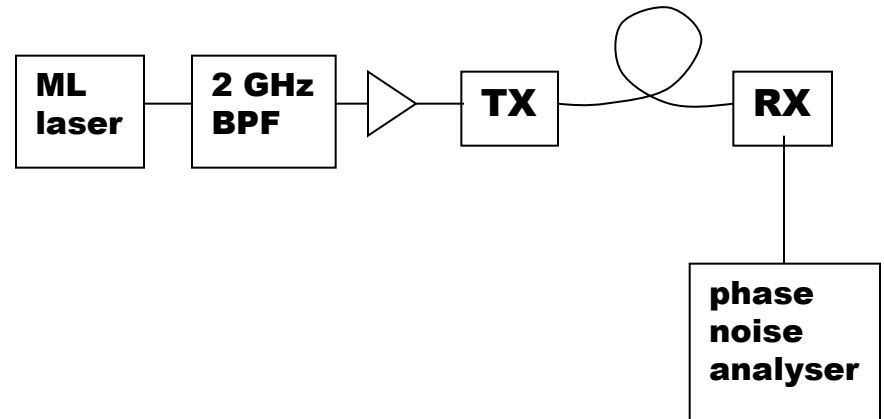
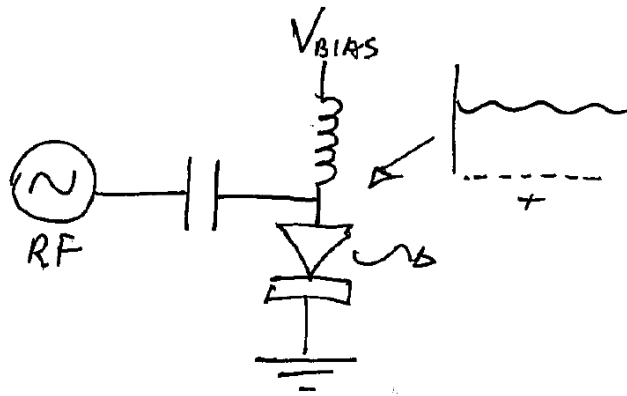


$$K_d = K_i + / - K$$

$K_i = 2\pi n_i / \lambda_0$ - wave vector of the incident beam.
 $K_d = 2\pi n_i / \lambda_d$ - wave vector of the diffracted beam.
 $K = 2\pi F / v$ - wave vector of the acoustic wave.

Direct modulation of laser diode

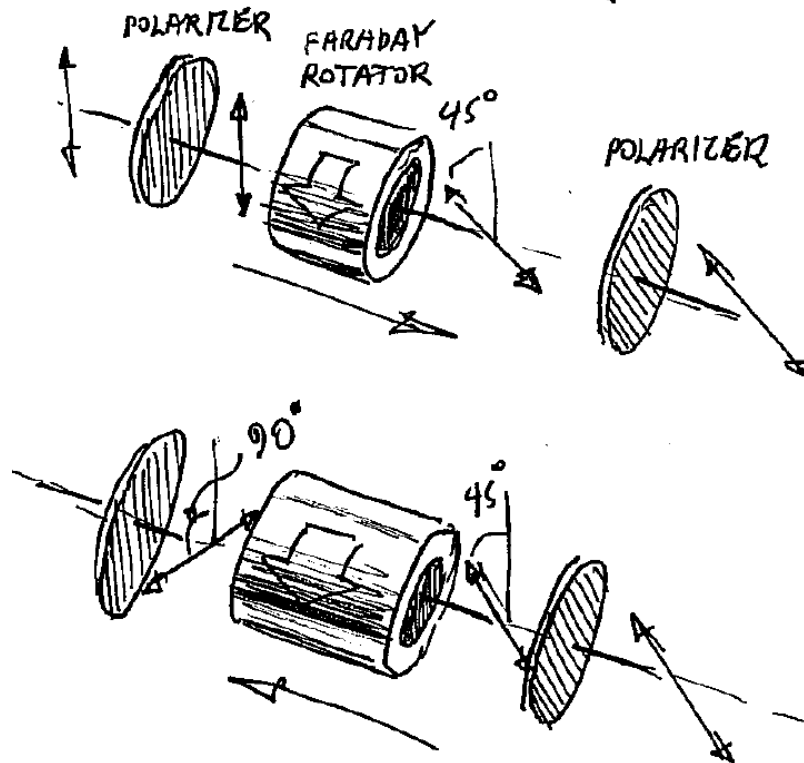
- **Common in telecom, cable TV**
- **Capable of very low noise**



- **Low noise RF source: modelocked laser pulse train on photodiode, bandpass filter at 2GHz**
- **Source jitter of 11 fsec RMS from 1kHz to 40MHz limited by photodiode shot noise at -155dBc**
- **Noise after link is 15fsec RMS, 1kHz to 40MHz**
- **No phase noise added by the link from 10Hz to 10kHz**

Isolator

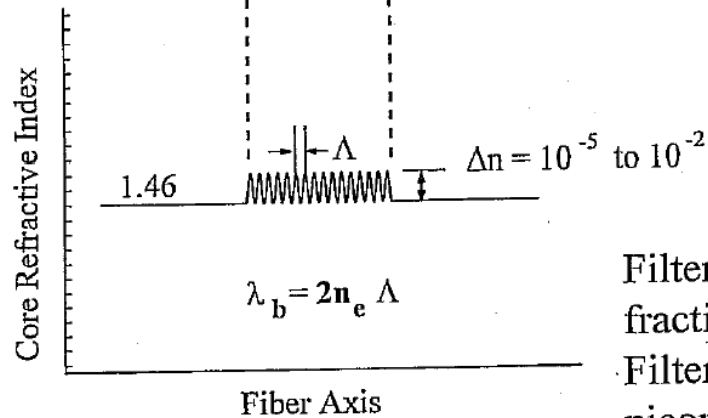
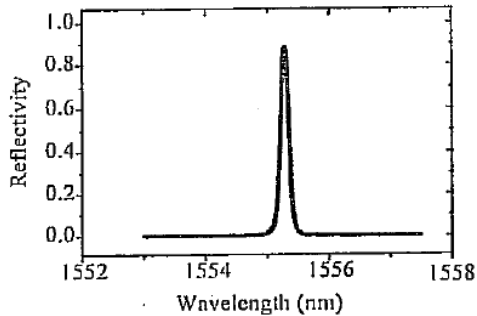
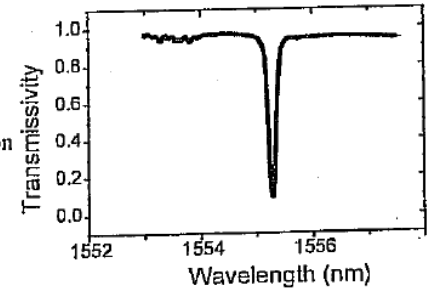
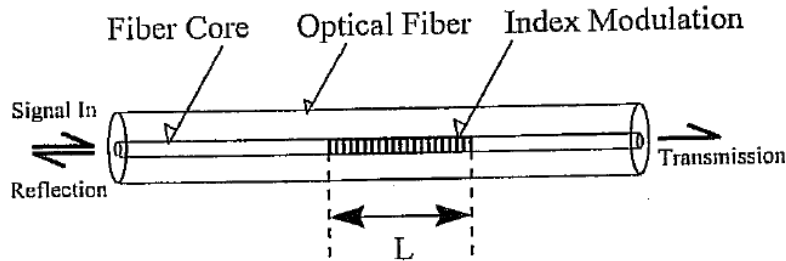
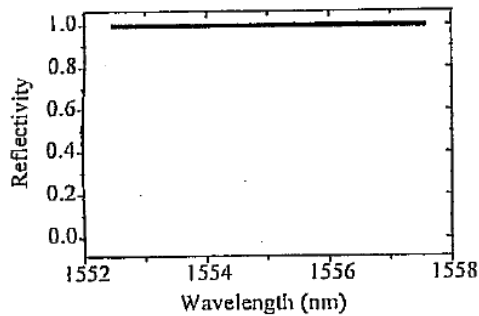
- Based on Faraday effect (RWV page 721)
- Non-reciprocal polarization rotation makes optical “diode” function
- Can be extended to make circulator
- Input polarization-independent versions created by splitting input into two arms, isolating each and then recombining



Filters

- **Dielectric bandpass or notch filter**
 - **Stack of transparent layers with alternating indices of refraction (RWV p. 295)**
 - **Direct analogy with periodic microwave filter designs**
- **Fiber Bragg grating**
 - **Periodic index modulation in fiber itself, very narrow band notch filter**
- **Birefringent filter (Lyot and Solc filters)**
 - **Based on wavelength dependence of polarization state after birefringent element**
- **Fabry-Perot resonator**
 - **Two-mirror resonator, tunable by varying spacing**
 - **Essentially one-dimensional fiber implementation has near ideal characteristics**

Bragg gratings in fiber

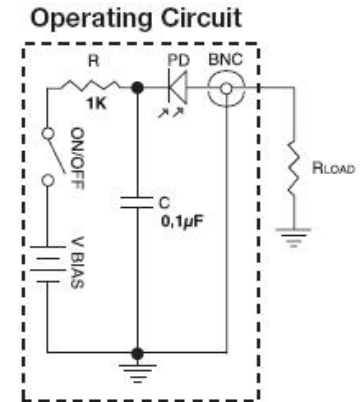


Filter Isolation:
fraction of a dB to >100dB
Filter Bandwidth
picometers to hundreds nm

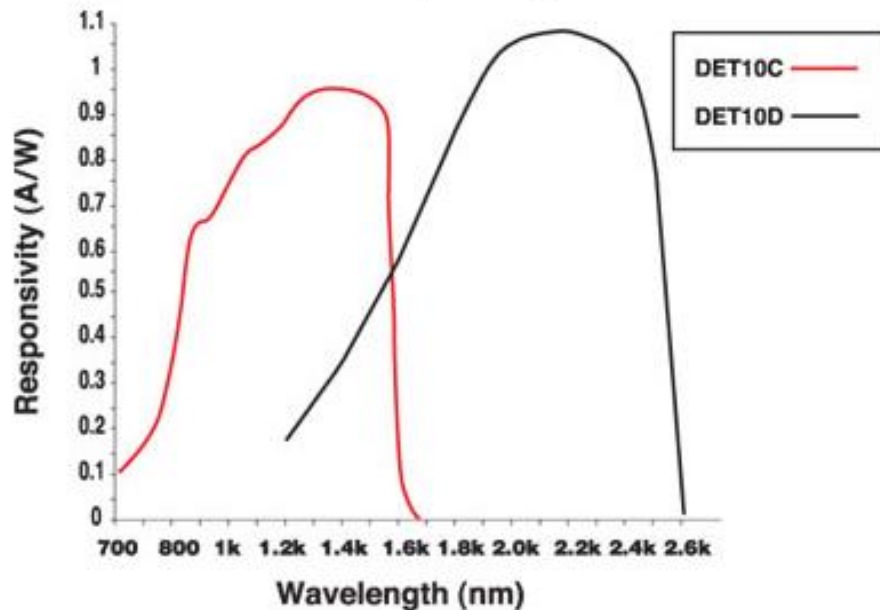
- **William Morey**

Detectors

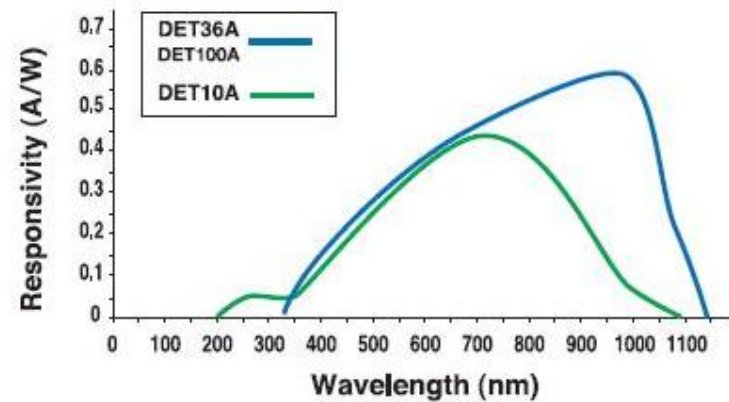
- **PIN junction photodiode is most common fast detector (~100GHz)**
 - **Bandwidth ~ 0.35/risetime for simple RC**
 - **Bandwidth ~0.44/pulse width for Gaussian**
- **Bandgap determines spectral sensitivity**



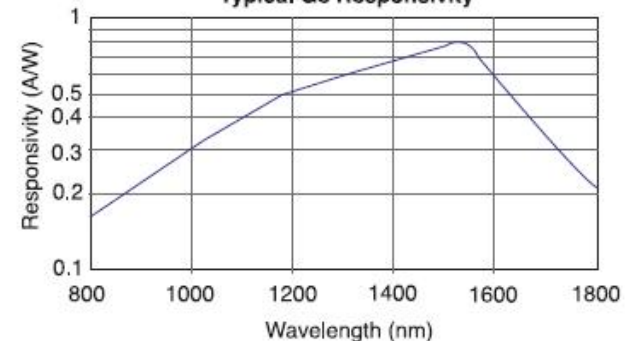
InGaAs Responsivity Curves



Silicon Responsivity Curves

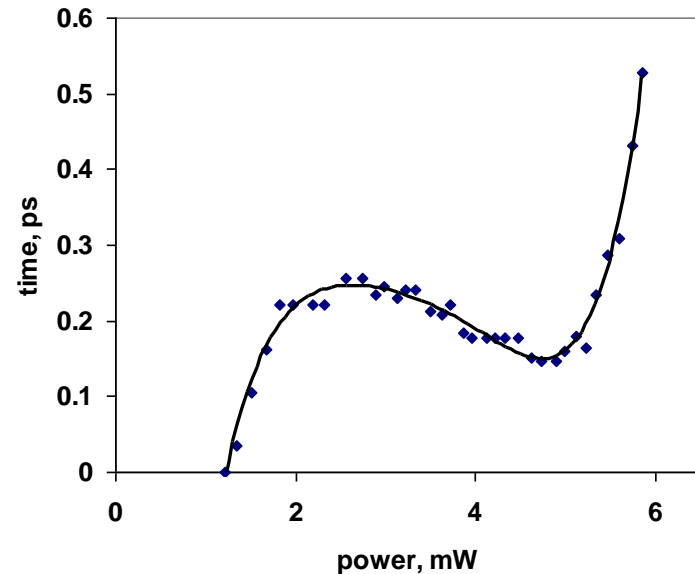
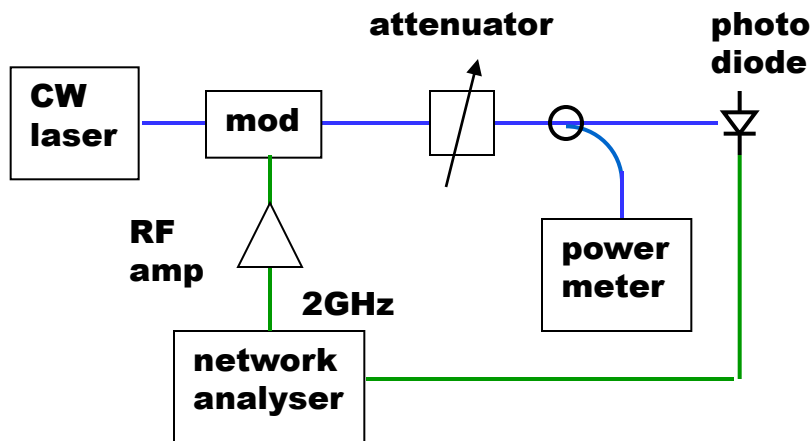


Typical Ge Responsivity



Measuring AM-to-PM characteristic

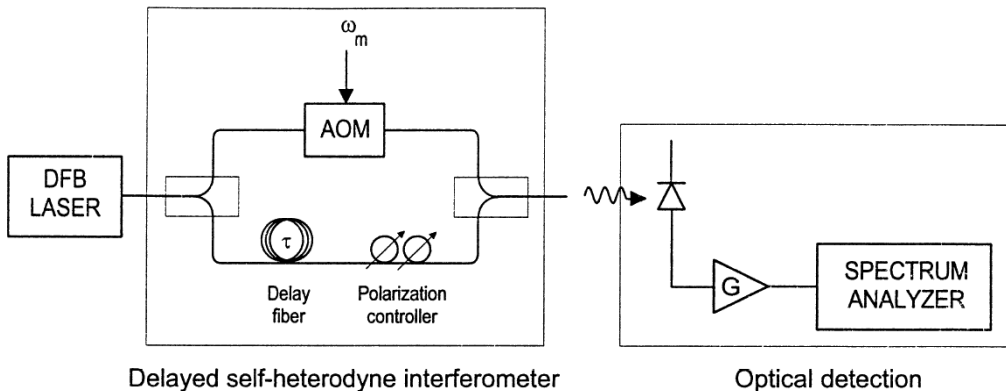
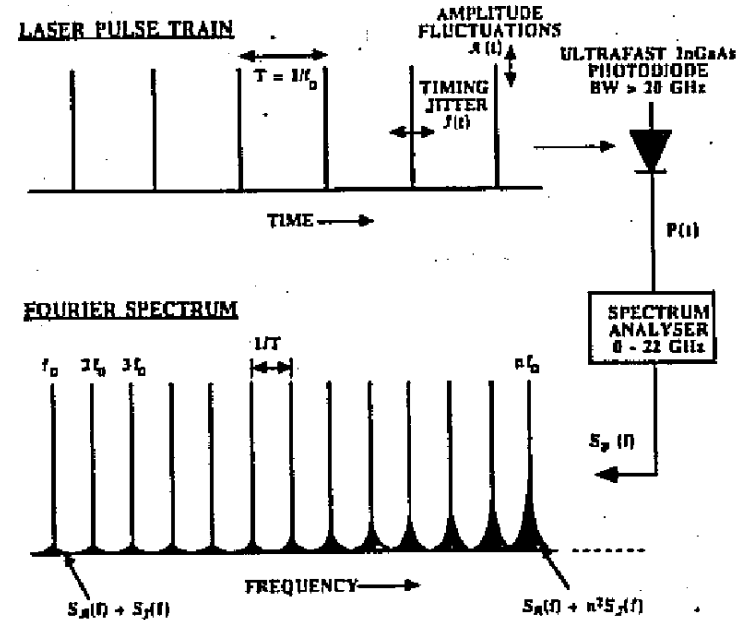
- **Saturation occurs when photocarrier density is high enough to diminish intrinsic electric field**
 - **Efficiency decreases, ultimately limiting signal amplitude**
 - **Junction capacitance is modulated, changing phase transfer function**
 - **“AM-to-PM”**
 - **Bandwidth decreases**
 - **Time domain response shows “tail”**
 - **Most diodes saturate at around 100-200mV into 50 Ohms, though higher sat power models are available**



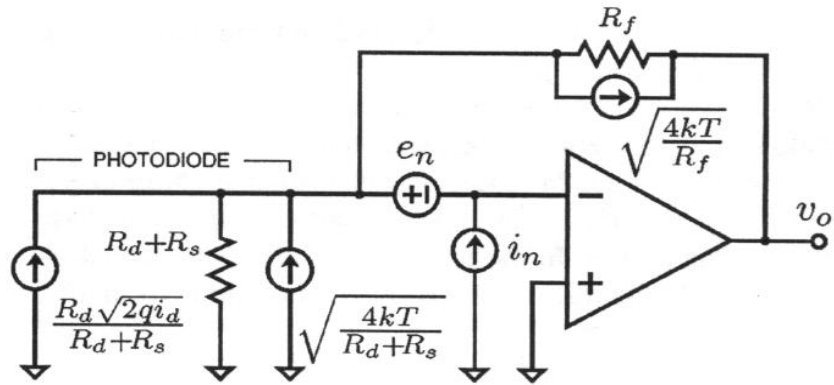
- **“Kink” in AM/PM curve provides zero-slope operating point**
 - **+/- 10% in power produces <10fs timing shift**
 - **Power is easily regulated to 1%**

Detection in frequency domain

- Note that photocurrent is proportional to optical power
 - Diode output is proportional to the square of the optical field ("square law detector"), thus can be used as a mixer to derive difference frequency (if it's within detector bandwidth)
 - Observe beats between two optical signals
 - Detect envelope modulation of optical wave



Noise sources in detection



$$\overline{v_o^2} = \left[\underbrace{\frac{2qi_d R_d^2}{(R_s + R_d)^2}}_{\text{DARK CURRENT SHOT NOISE}} + \underbrace{\frac{4kT}{R_s + R_d}}_{\text{PHOTODIODE JOHNSON NOISE}} + \underbrace{i_n^2}_{\text{OP-AMP CURRENT NOISE}} + \underbrace{\frac{4kT}{R_f}}_{\text{FEEDBACK RESISTOR JOHNSON NOISE}} \right] R_f^2 + \underbrace{e_n^2 \left(1 + \frac{R_f}{R_s + R_d}\right)^2}_{\text{OP-AMP VOLTAGE NOISE}} \quad \text{V}^2/\text{Hz} \quad (28)$$

where

- i_d dark current;
- i_n op-amp input noise current spectral density;
- e_n op-amp input noise voltage spectral density;
- R_d photodiode dynamic resistance;
- R_s photodiode series resistance.

